



OBFN activities within Thales Photonic Lab

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M. Maignan, M. Sotom,... Thales Alenia Space



- **Introduction**
- **Early experiment**
- **Recent achievements**
 - ▶ Ultra-Wideband free-space beamforming
 - ▶ Beamforming for ground radar
 - ▶ Beamforming for communications
- **Advanced studies :**
 - ▶ OBFN concepts : adapted LO as an example
 - ▶ OBFN components :
 - Tunable delays : PBG, SOA
 - Switches : new EO materials
 - Dual-frequency / SSB modulation schemes : DFL



Thales Alenia Space

satellite applications

Thales Comm.

Mil Com applications

Thales Laser

solid state lasers

Thales Netherlands

naval applications

Thales Sensors UK

naval ESM appli.

Alcatel-Thales 3-5 Lab.

optoelectronic components

Thales Aerospace Div.

- RF modules and optical links developments
- Photonic architectures in airborne systems

Thales Air Systems Div

- RF & Digital optical link developments
- Photonic architectures for ground based radars

Thales Research & Techno.

- photonic architectures
- optoelectronic processing

TTCS

- photonic architectures for PAA

Thales@NTU

- photonic components

Thales Photonics
BGs-Corporate Lab.

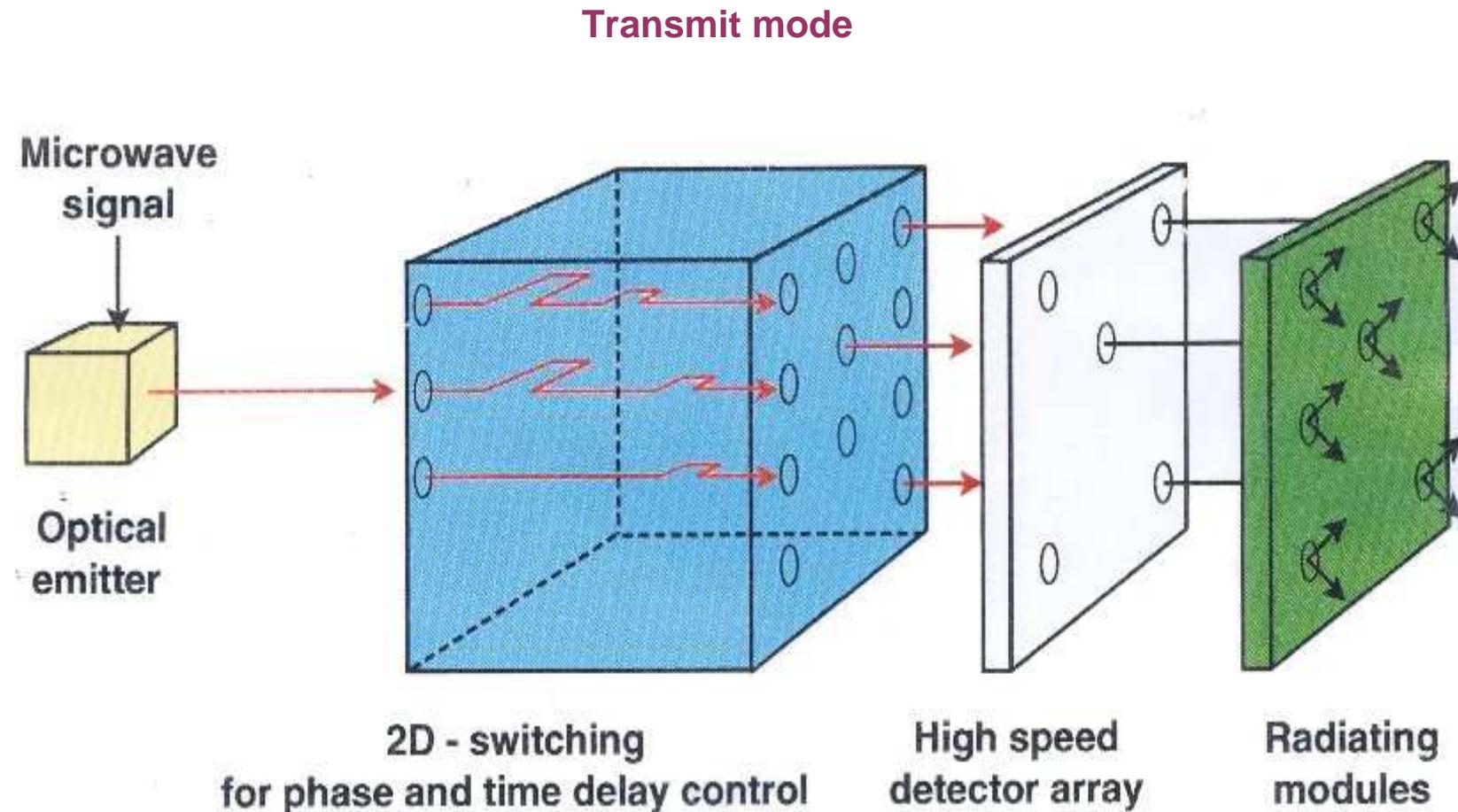
Research Institutes

- Wideband RF Summation
- Switching matrices
- Photo HBT & mixers
- Dual frequency sources....

SMEs

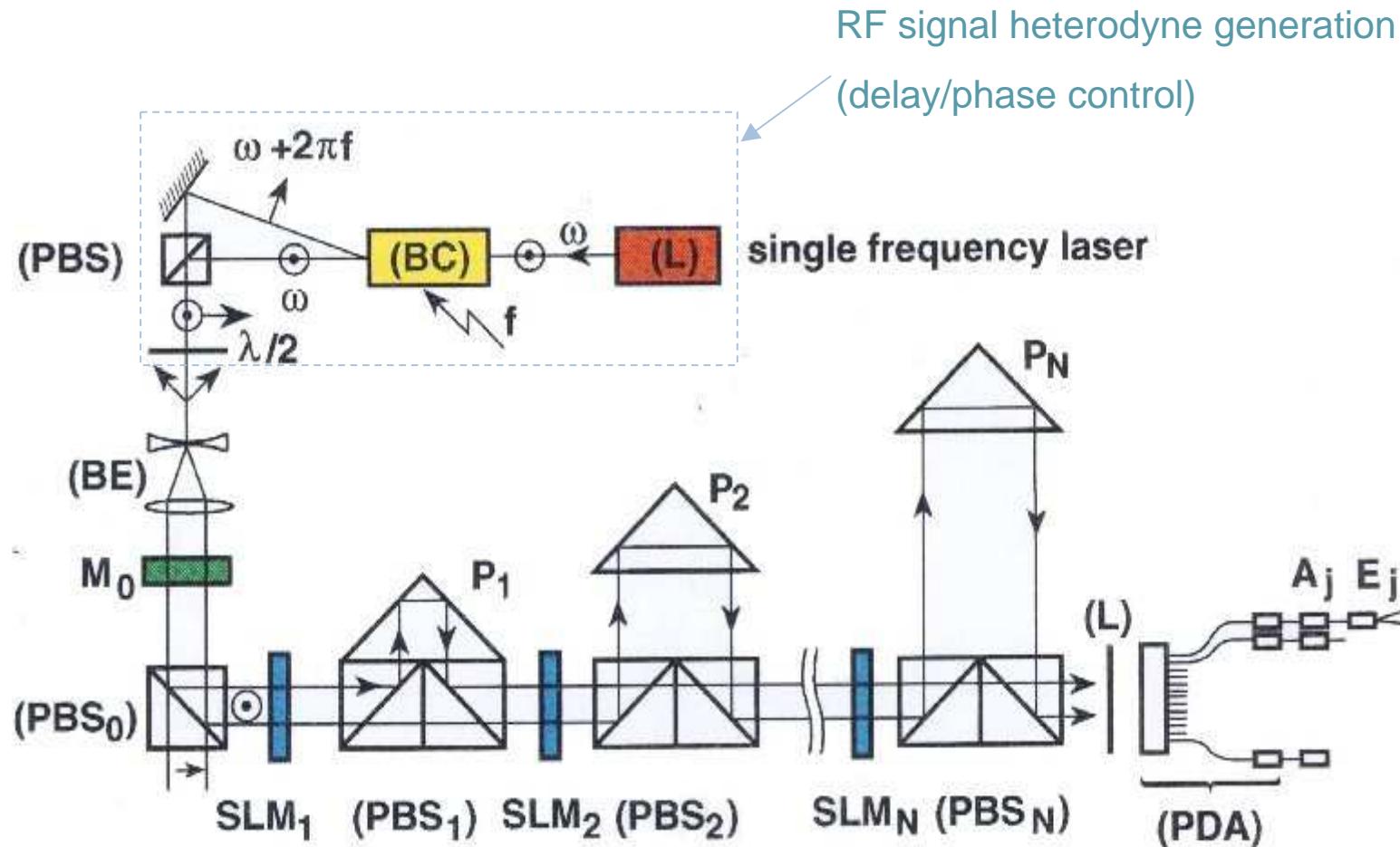
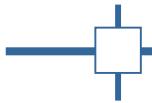
- RF optical modules
- Integrated WDM modules
- Optical waveguides
- ...

Optically controlled phased array antennas



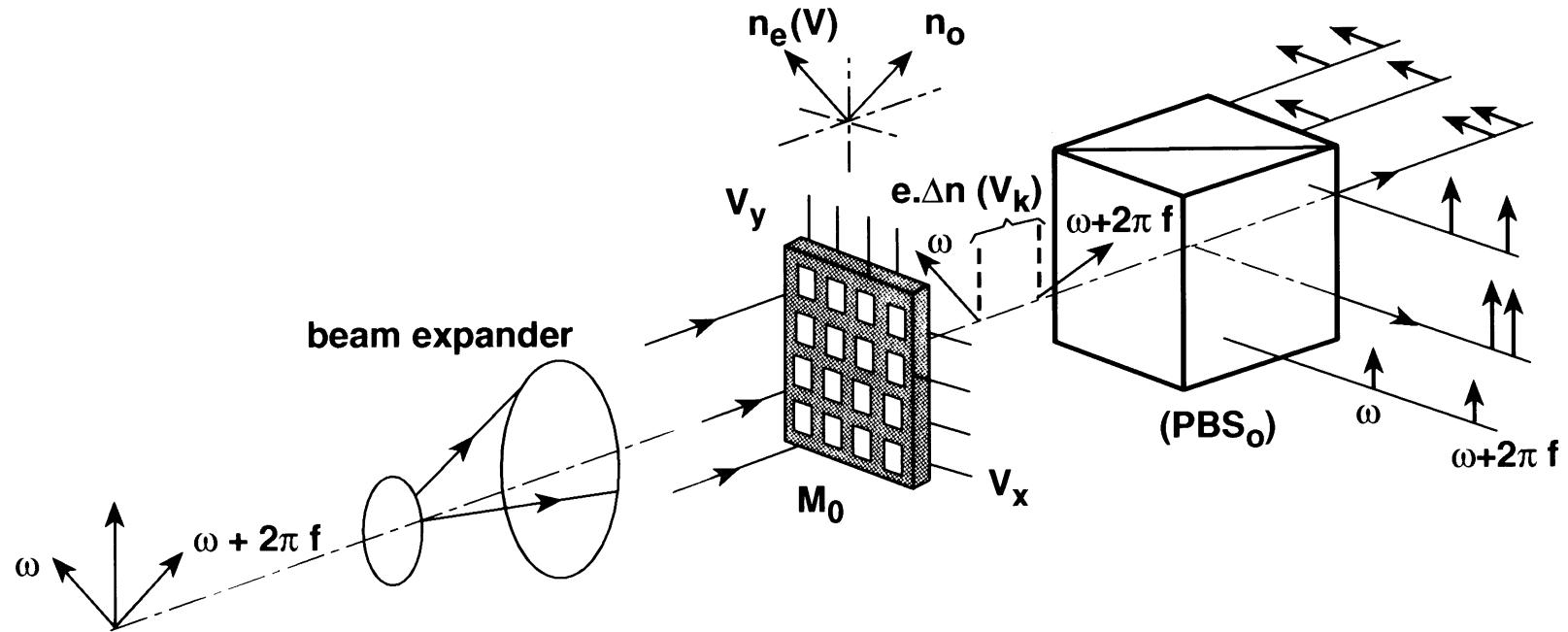


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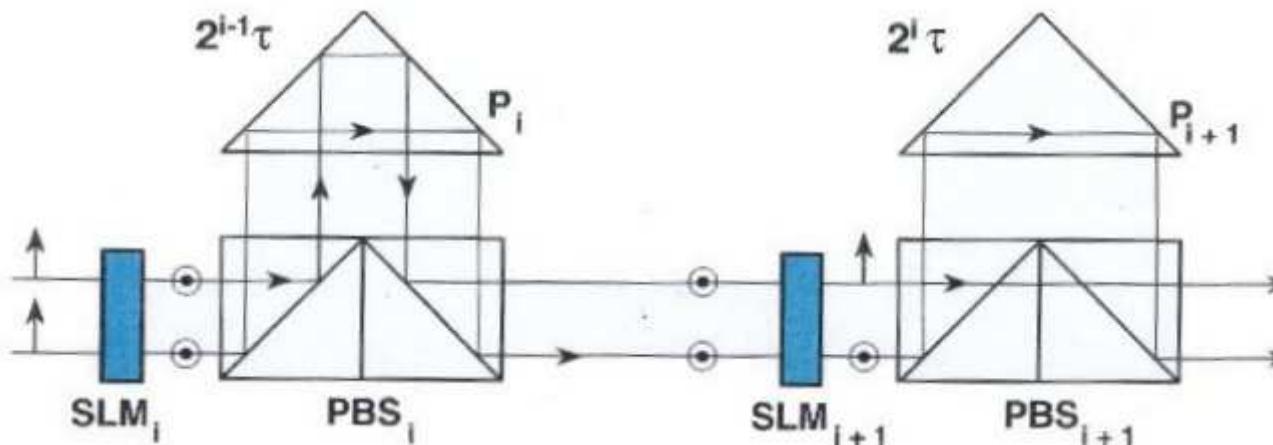


SLM_i : polarization rotation

Implementation of the phase control

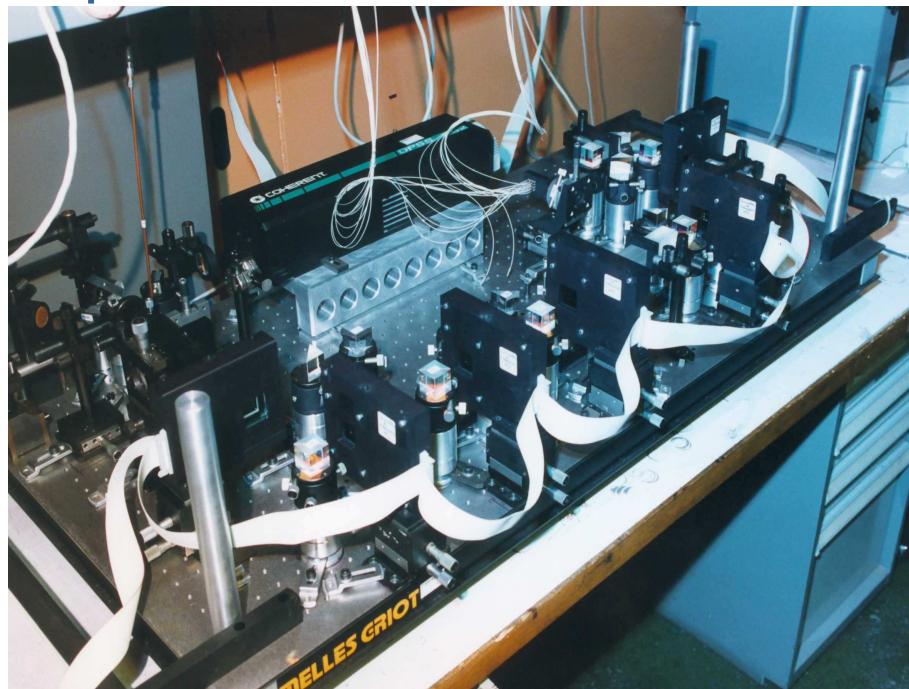


- Electrically controlled birefringent mode
- $p \times p$ pixels
- on channel k : $i(k) = i_0 \cos(2\pi f t + 2\pi e \Delta n(v_k)/\lambda)$ with $\Delta n(v_k) = n(v_k) - n_0$



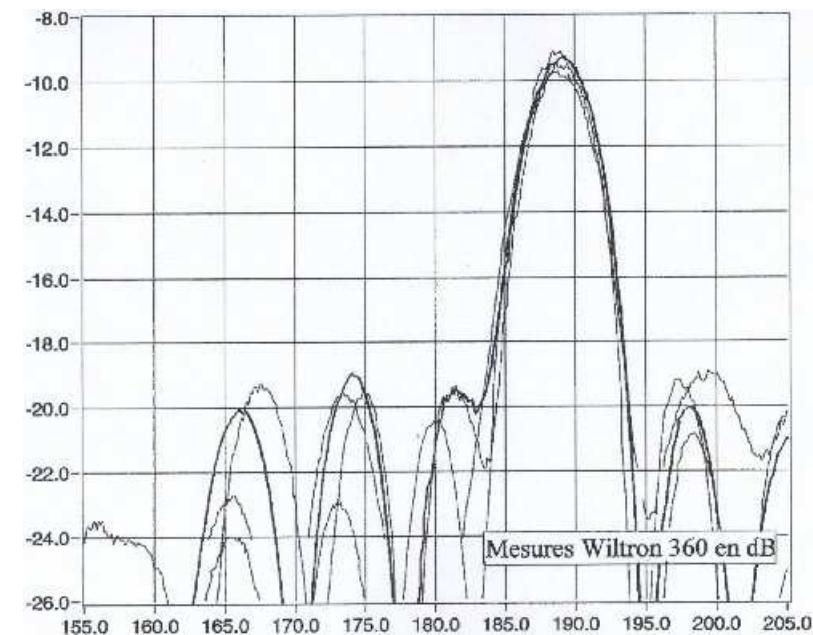
- N SLMs $\Rightarrow 2^N$ delays
 - each SLM_i : $p \times p$ pixels
 - channel k : $i_k(t) = i_0 \cos(2\pi f t + 2\pi f \sum \epsilon_{kj} 2^{j-1} \tau)$
 - If reflexion on PBS_j $\epsilon_{kj} = 1$
 - Else $\epsilon_{kj} = 0$

Free-space optical TTD for S band antenna scanning



- delays: 5 bits
- phase control: 6 bits
- 16 channels, 16 radiating elements
- antenna beam pattern:
 - scan angle : $\pm 20^\circ$
 - BW : 2700 - 3100 MHz
 - no beam squint

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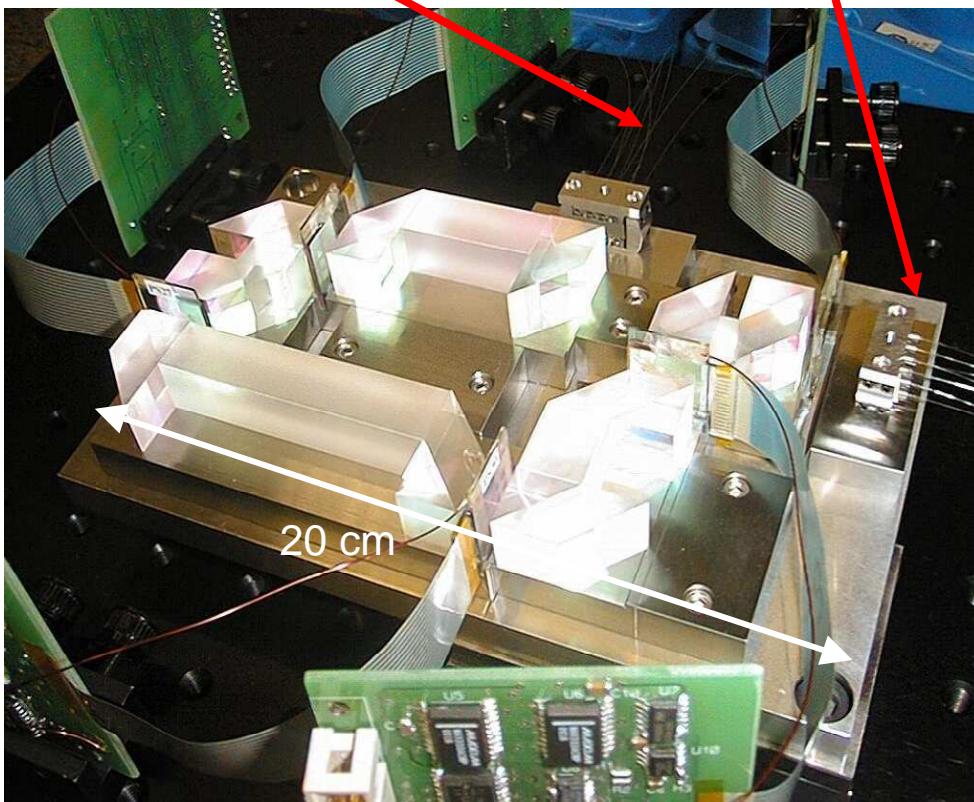


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Compact True Time Delay module for transmit and receive modes

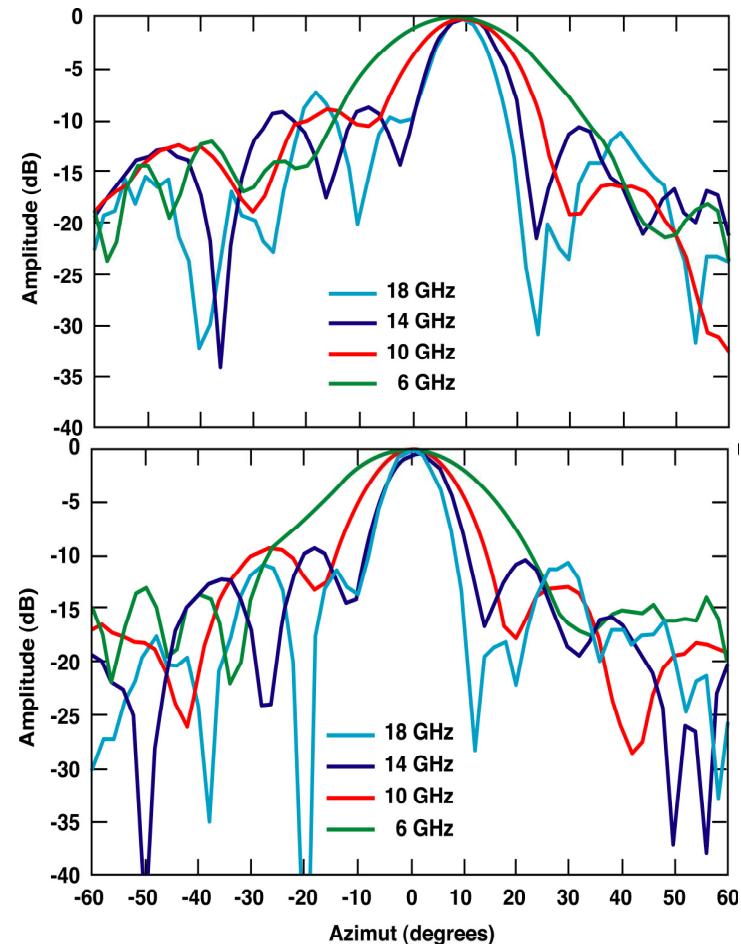
output fibres

input fibres



- BW = 2-20GHz
- 8 channels, 8 radiating elts
- unit delay $\tau=6.5\text{ps}$
- 5 SLMs → 32 delays/ch.
- $t_{\text{on}}=20\text{ms}$, $t_{\text{off}}=100\text{ms}$

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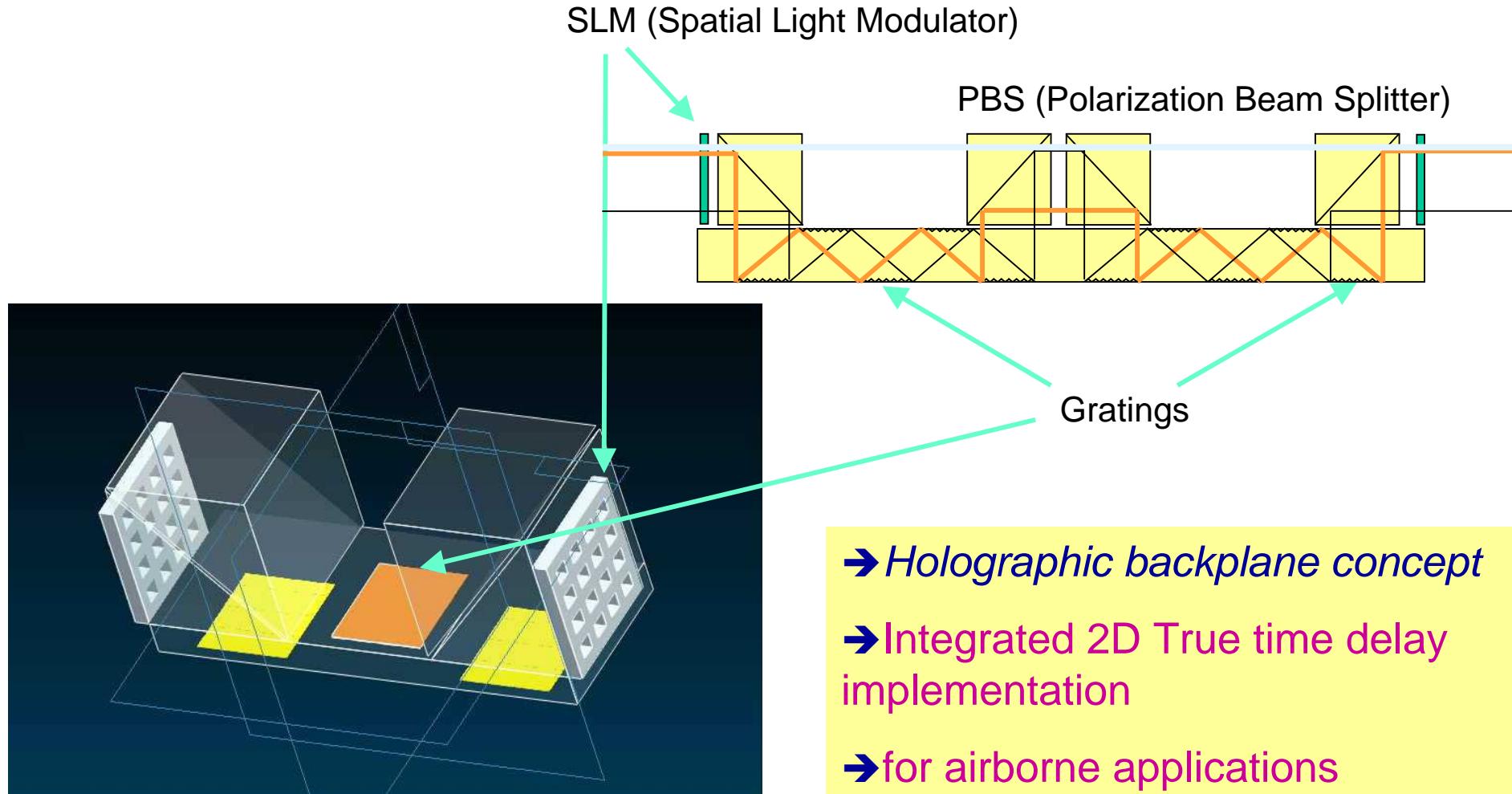


**measured far field pattern
for:**

- scan angle : $\pm 20^\circ$
- frequency : 6 – 18 GHz
- no beam squint

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Advanced concept for ultra-compact architecture

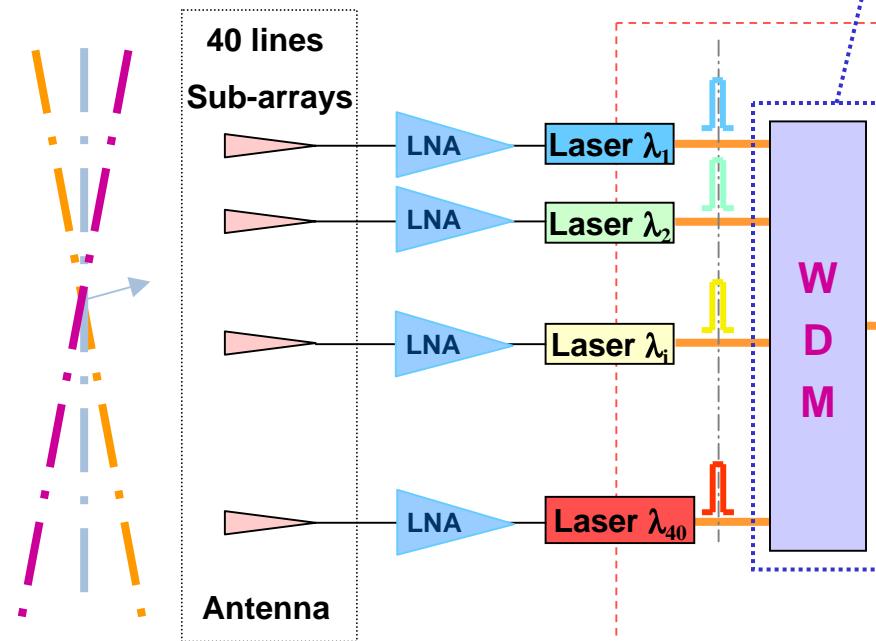




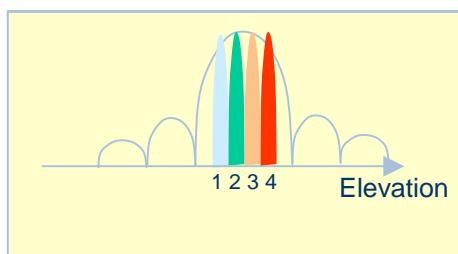
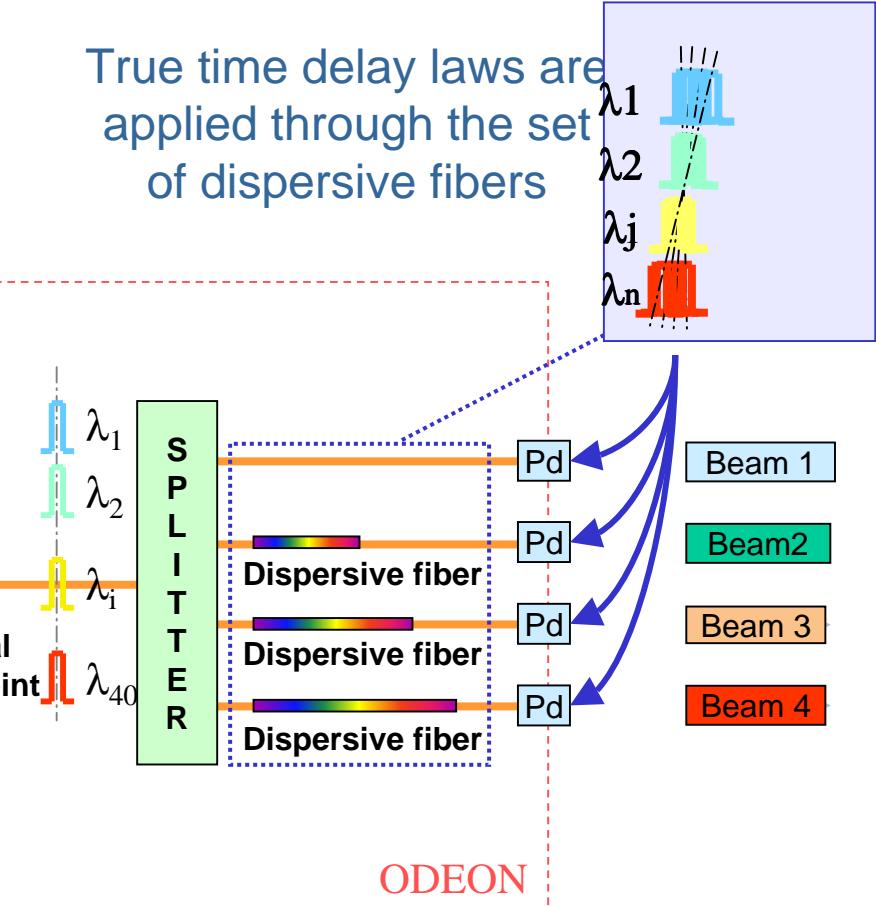
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Rx Dispersive OBFN principle

Optical summation
of RF signals



True time delay laws are applied through the set of dispersive fibers

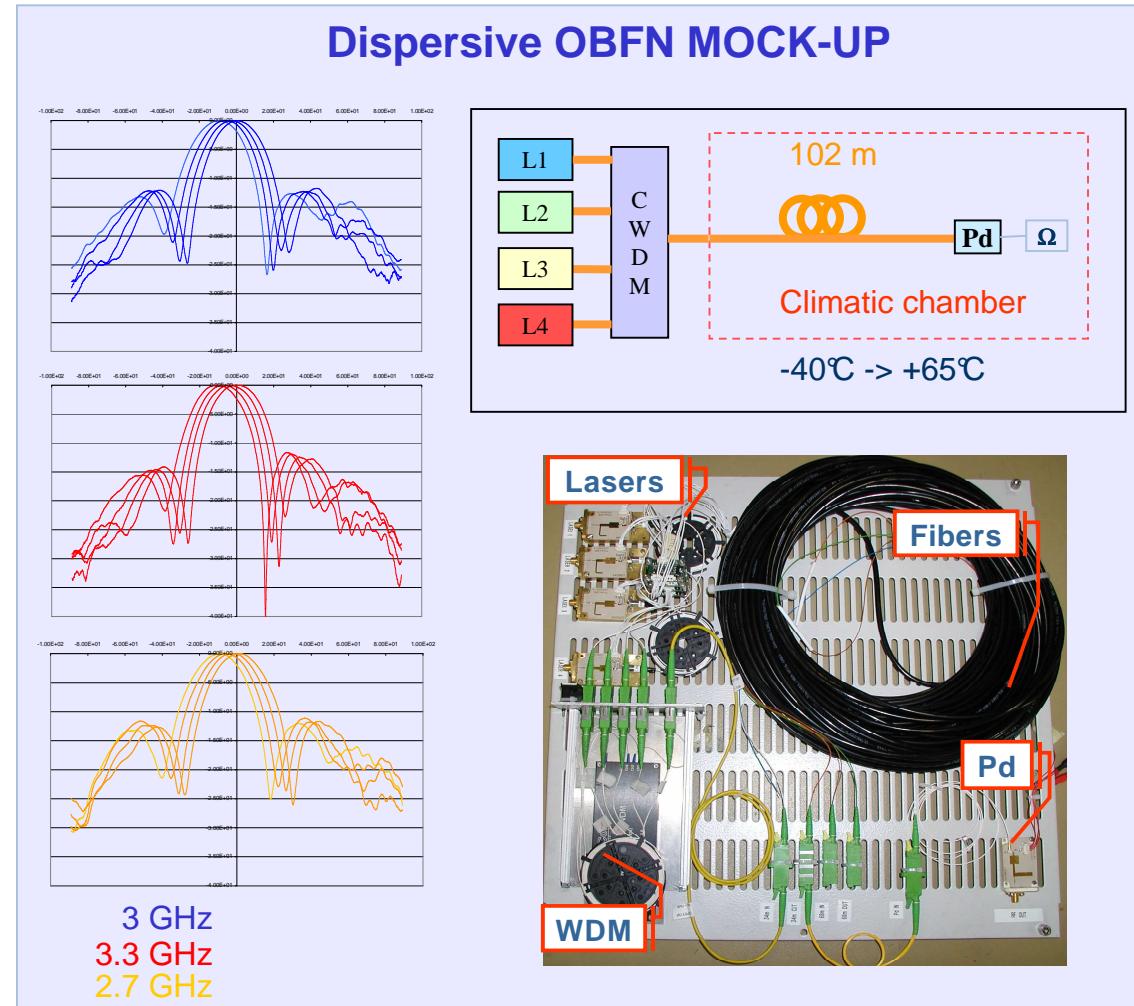
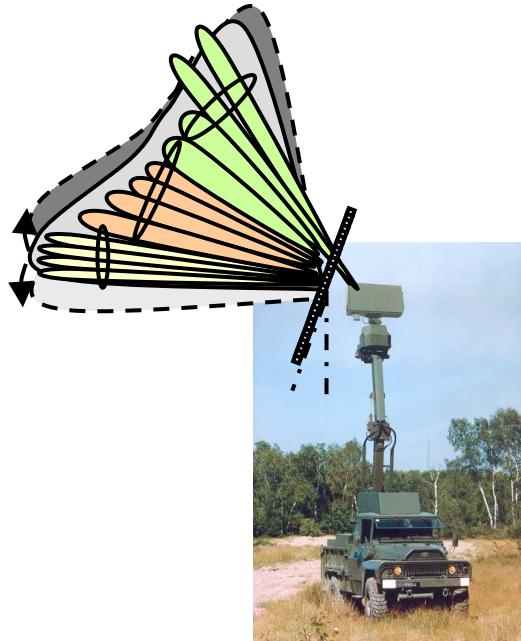


LNA : Low Noise Amplifier
WDM : Wavelength Division Multiplexing
Splitter : standard balanced optical power splitter

Dispersive OBFN

principle has been validated with a 4 channel mock-up

- Radiating patterns without beam squint
- No temperature sensitivity



Several OBFN architectures have reached TRL 4 mainly depending on operating bandwidth



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Within SATn'LIGHT :

- ▶ Optically Controlled Antenna demonstrator
- ▶ Building blocks overview
 - ▶ Dual-Frequency laser
 - ▶ Single-polarization Modulator
 - ▶ Phase & Amplitude Spatial Light Modulators
 - ▶ Optical summation module



Within SAT ‘N LIGHT we have selected a FAFR (Focal Array Fed Reflector) Rx antenna as a “study case”.

- well-suited for narrow angular coverage for GEO satellite
- based on cheap radiating aperture of a reflector
- many clusters of horns form numerous beams



Ka band, 1 GHz bandwidth, diameter = 1.2 m, focal length = 1 m

170 feeds for 34 beams in Northern Hemisphere

Why to optically steer an active antenna ?



Beam forming is a complex architecture from 170 RF inputs towards 34 outputs with

- ≈ 400 RF amplitude controls
- ≈ 400 RF phase controls
- ≈ 200 splitters / combiners

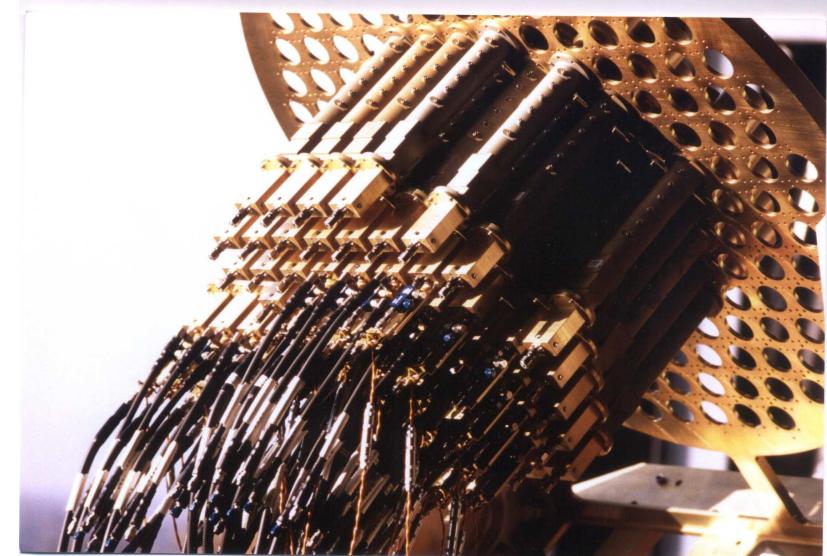
Optical architectures lead to drastic weight and volume savings.

Among the 2 classes of optical architectures :

- True Time Delay Beam Steering
- “Coherent” Phase Steering

We have selected the latter approach since it corresponds to the ultimate integration capacity for satellite

application



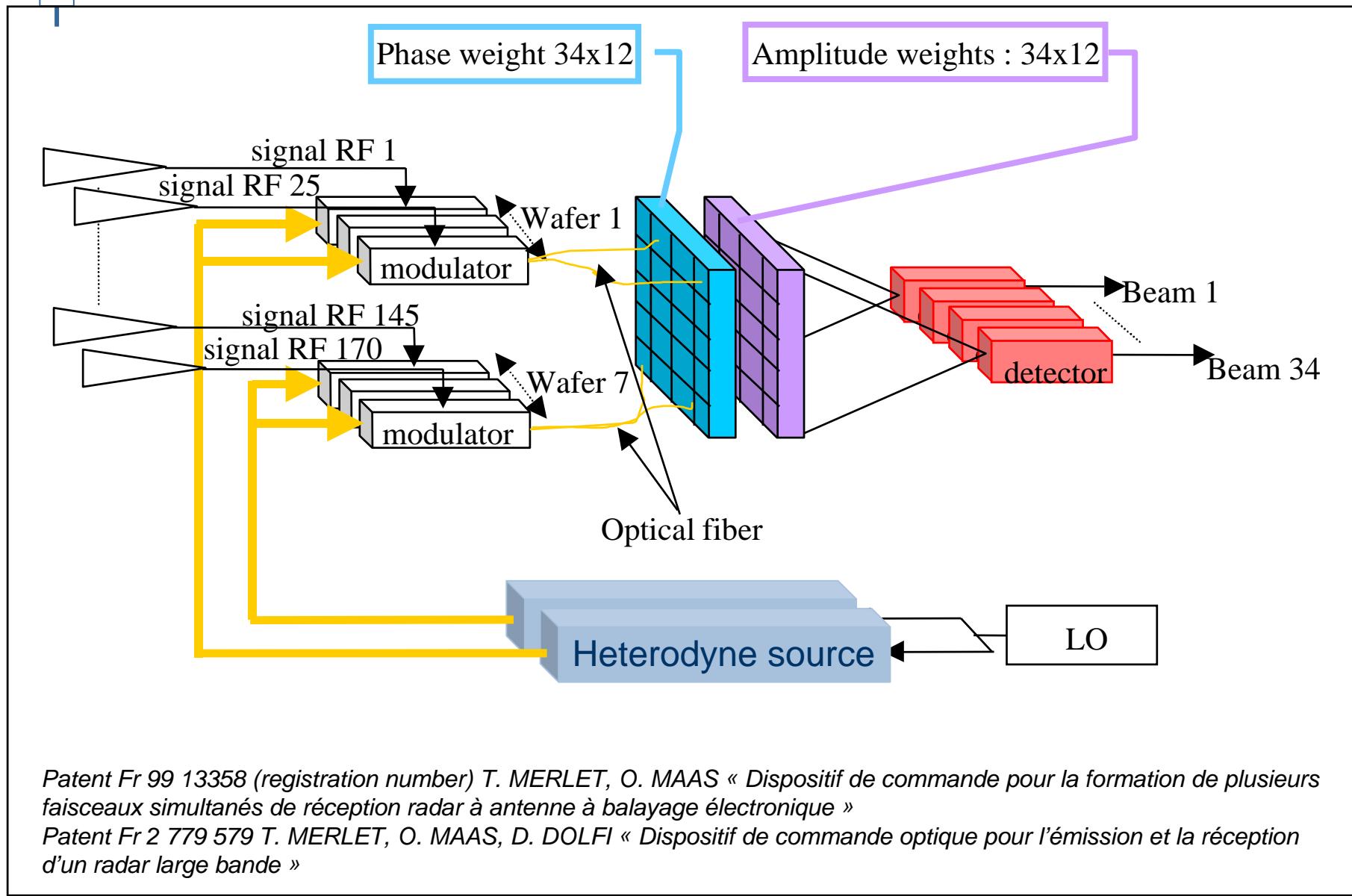
Coherent Optical Controlled Antenna (OCA) estimation* :

- Mass < 50 kg
- Consumption < 220 W
- Volume < 6 liters

* From SAT N LIGHT phase 1

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FAFR implementation of OCA architecture



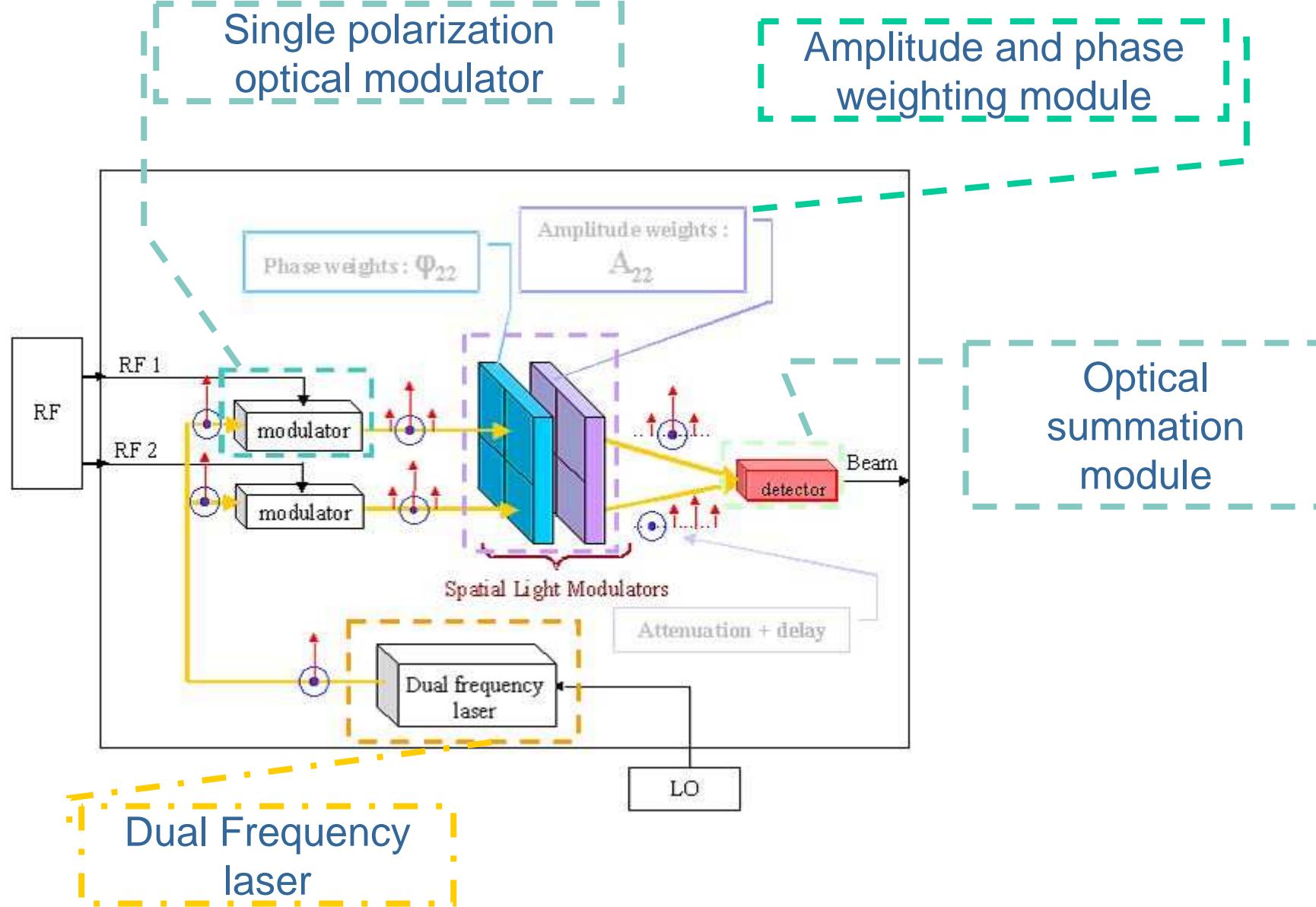


Optically-controlled antenna demo

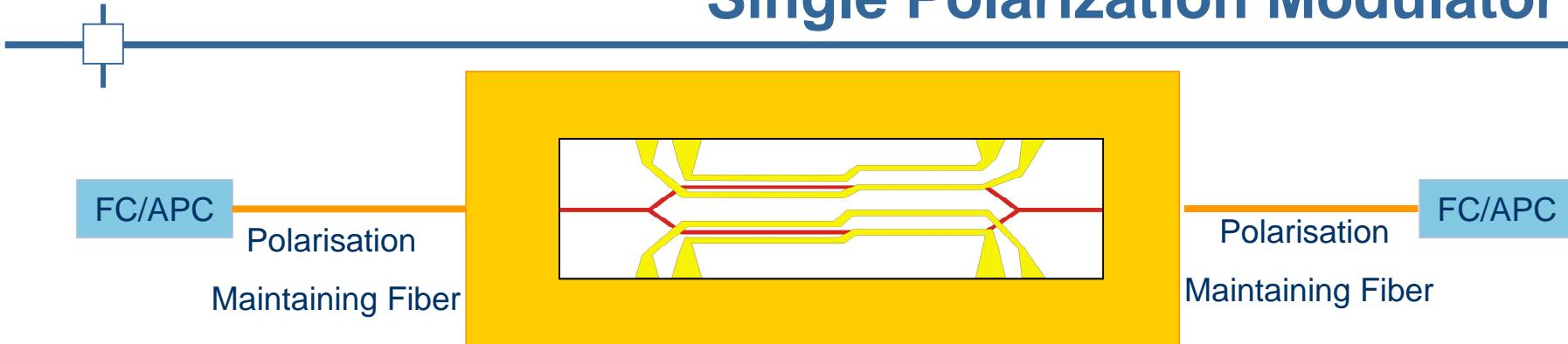


- Optically Controlled Active antenna (OCA) is a **proof of concept** demonstrator based on a coherent optical architecture
 - Heterodyne LO generation owing to cross-polarised modes in a solid-state cavity. *New concept*
 - RF Phase modulation on a single-polarisation *New concept*
 - Parallel 2D matrix for phase and amplitude weightings
 - Optical summation and optical Mixing on a single device *New concept*
- OCA demonstrator is composed of 4 building blocks
 - Dual-Frequency Laser
 - Single-polarization optical modulator
 - Amplitude and phase weighting module
 - Optical Summation module

Optically-controlled antenna demo



Single Polarization Modulator



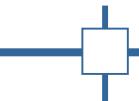
Coherent architectures are based upon delay coding in the range of 1 optical wavelength ($1 \mu\text{m}$).

- For RF phase stability, optical beams shall witness the same variations.
- Beams need to be carried on superposed on the same media

“Single polarization modulator” means :

- A modulator that transmit both polarizations, contrary to usual modulators.
- A modulator that only modulates one of the polarization.

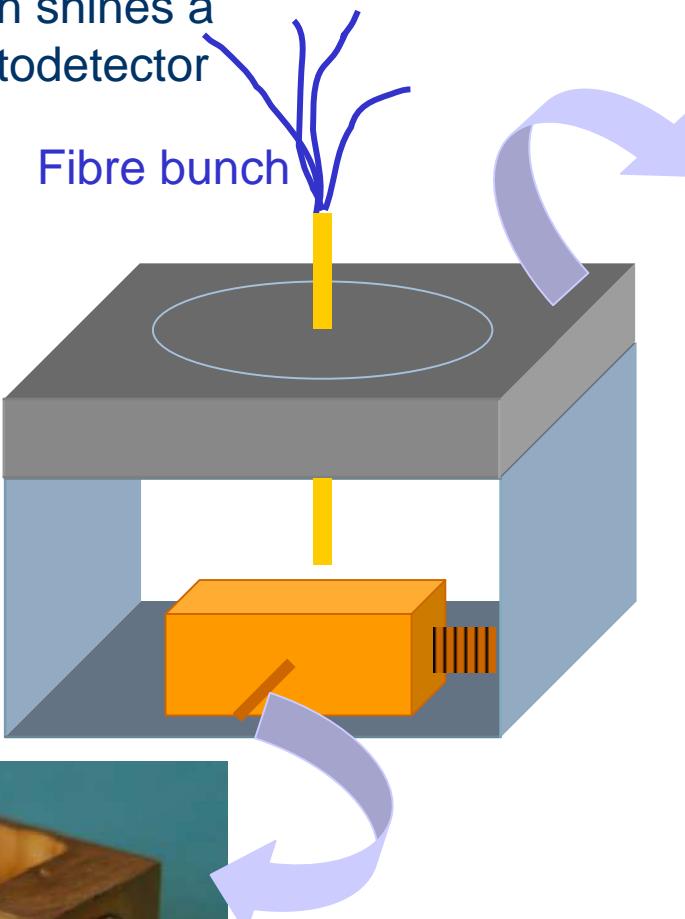
Refer to Photline presentation for detailed explanations



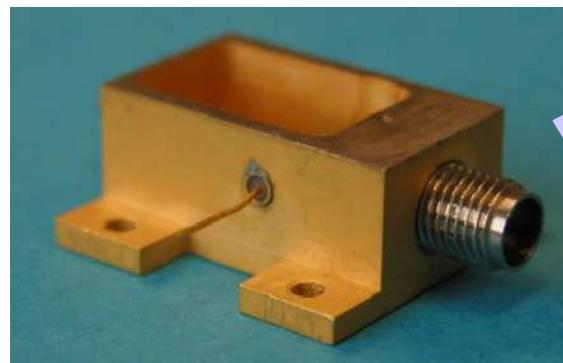
Summation module : manufacturing



8 fibers gather in a bunch shines a single large surface photodetector



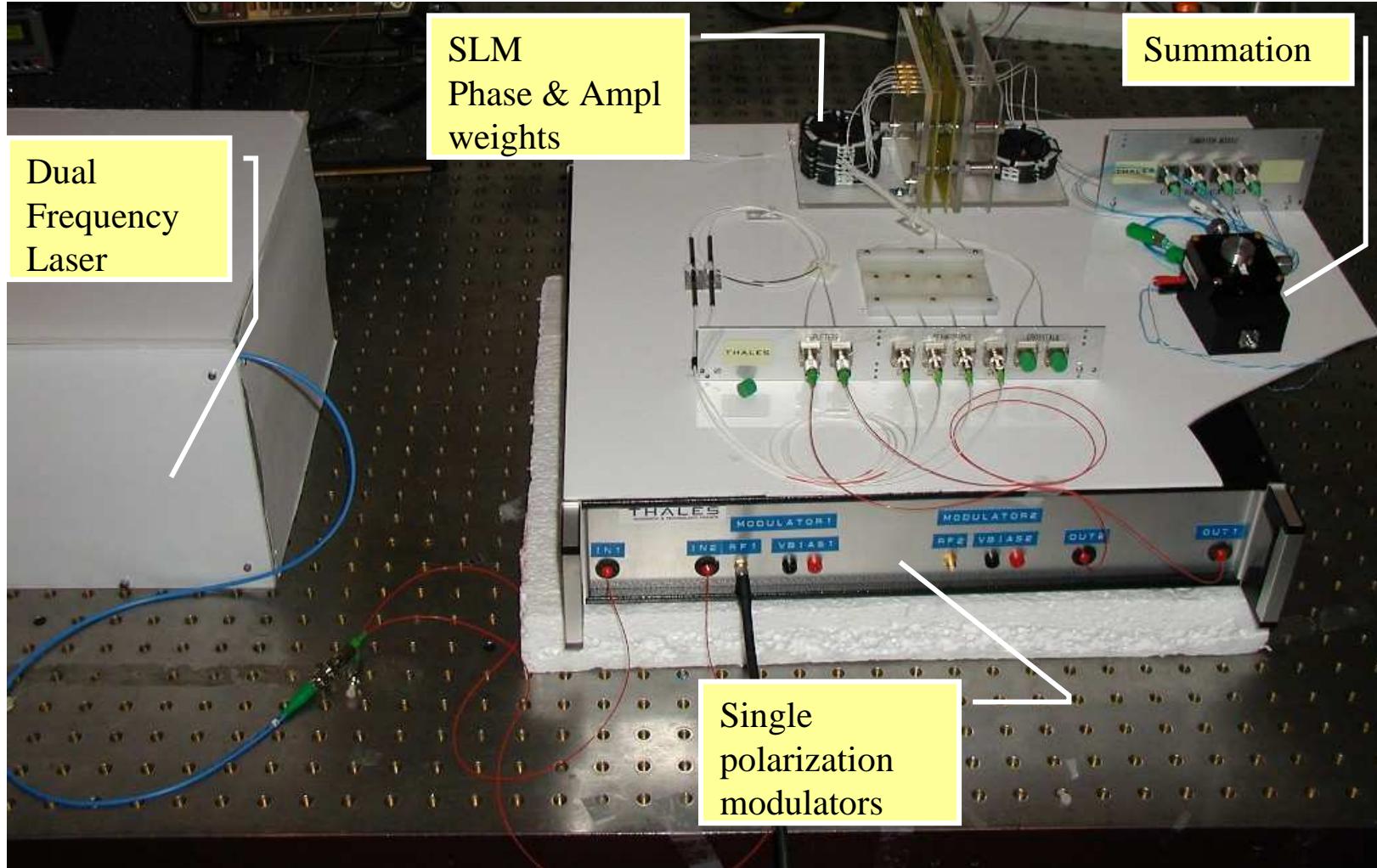
XYZ positioner to adjust optical fibre position before hermetic sealing



Box with:

- integrated PIN detector
- MW output (and DC bias)

OCA Integrated Demonstrator





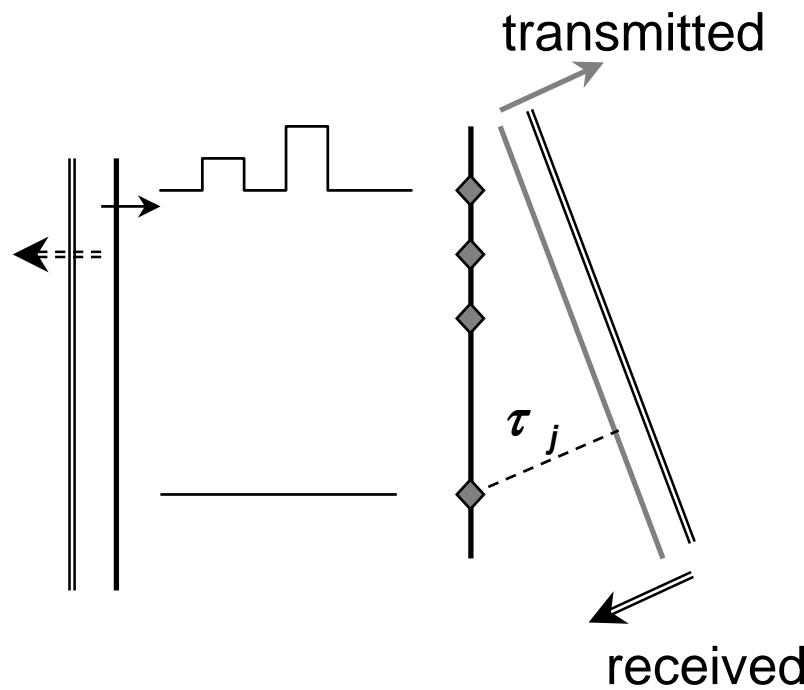
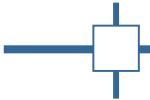
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Analog processing of the receive mode

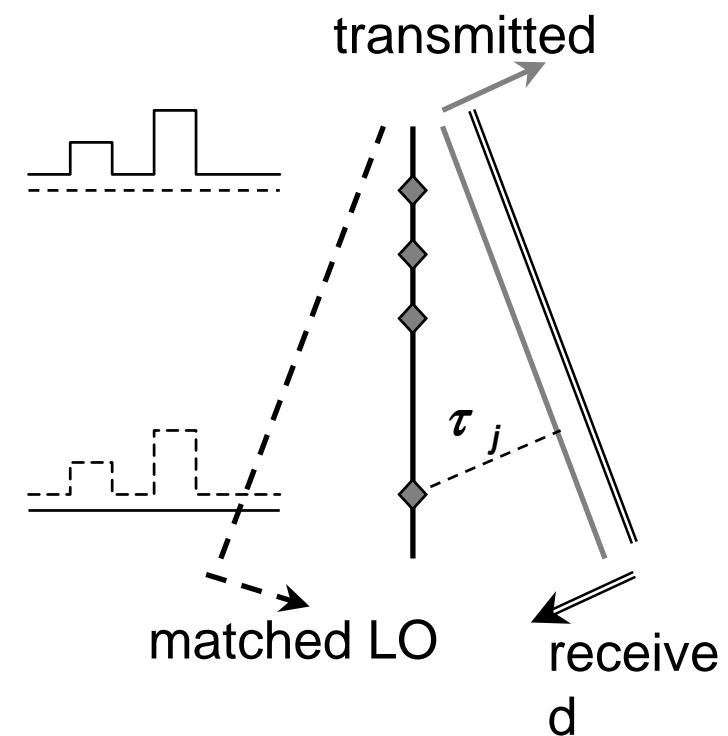


- microwave signals spread over a large dynamic range (up to 100 dB)
 - frequency bandwidth from <10% (radar, comm) up to 2-18 GHz (E.W)
 - in-phase addition over a large frequency BW of the received signals with a precision of few degrees
 - ➔ need of pxp analog microwave / optical links with large dynamic range and low noise figure
 - ➔ limited dynamic range of opto-links (typ. 90-95 dB in 1 MHz BW)
- *architecture with matched local oscillator (**time-delayed and optically carried**)*

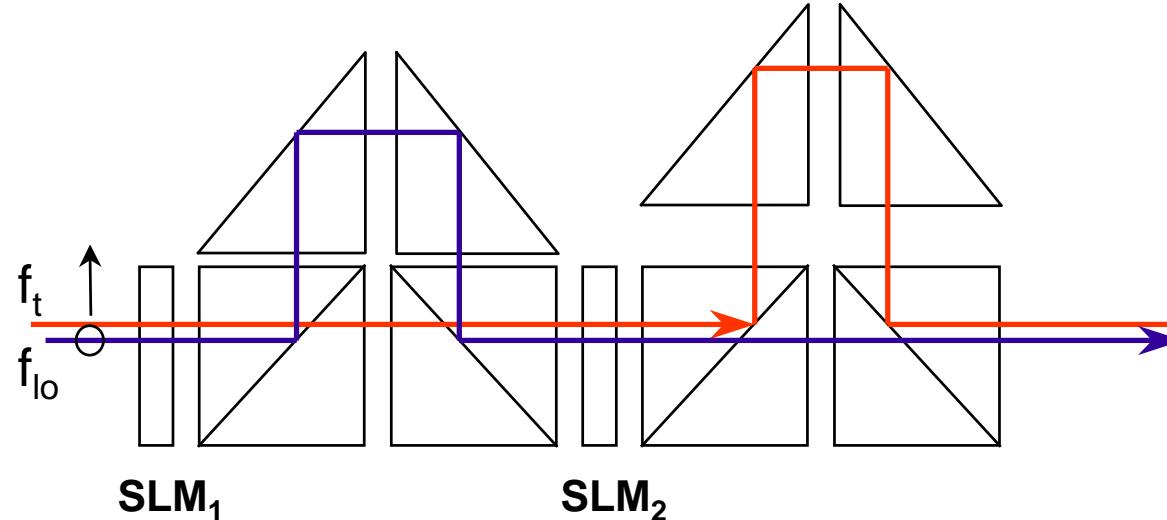
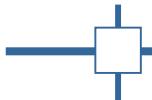
Time reversal vs matched L.O



direct transposition → time reversal

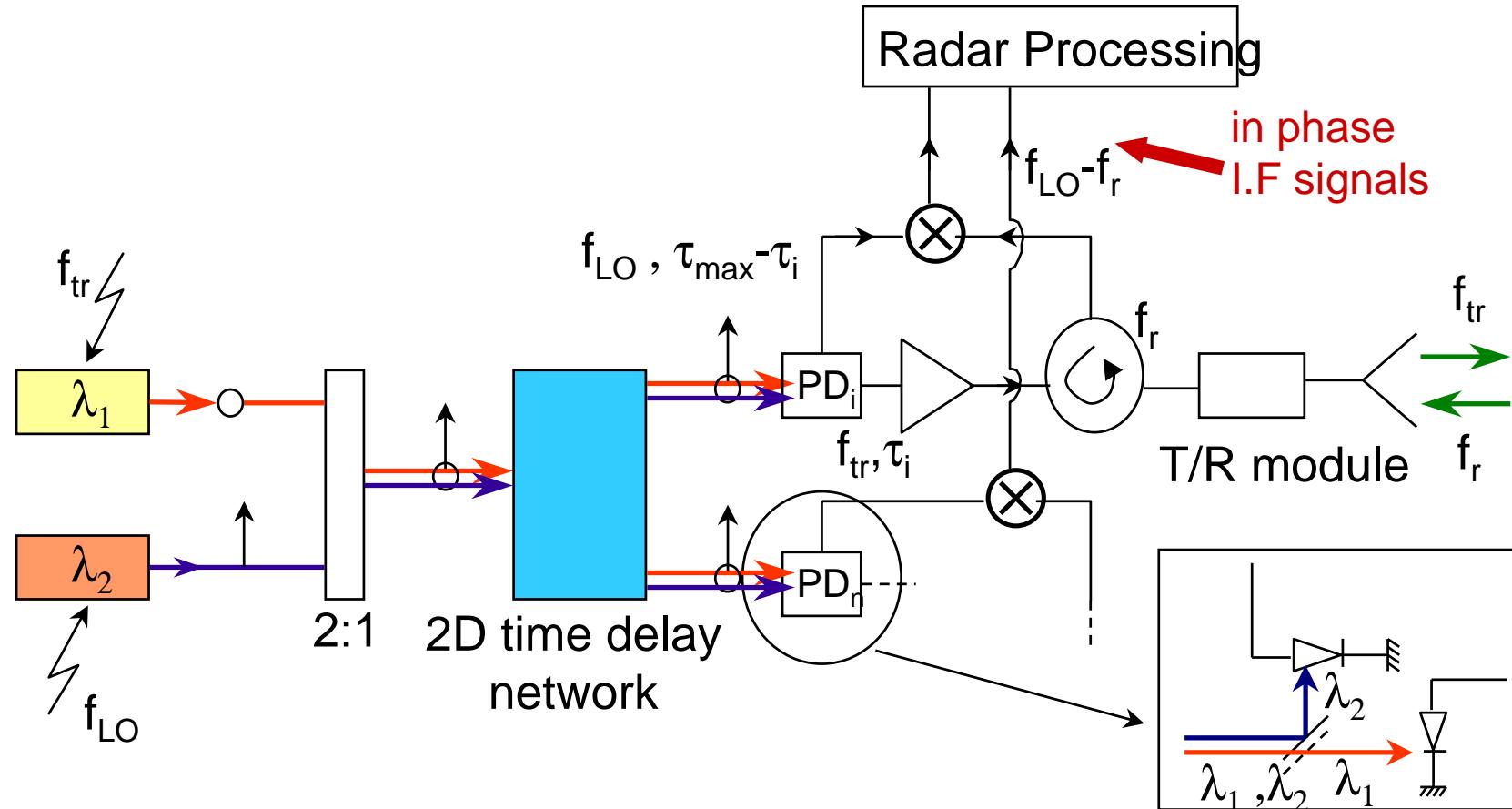


matched L.O → phase conjugation



- optical carriers of the transmitted signal and of the L.O
 - with crossed polarizations
 - at different wavelengths
- complementary time delays on each channel i :
 - $\tau_{\max} - \tau_i$ for the matched L.O
 - τ_i for the transmitted signal

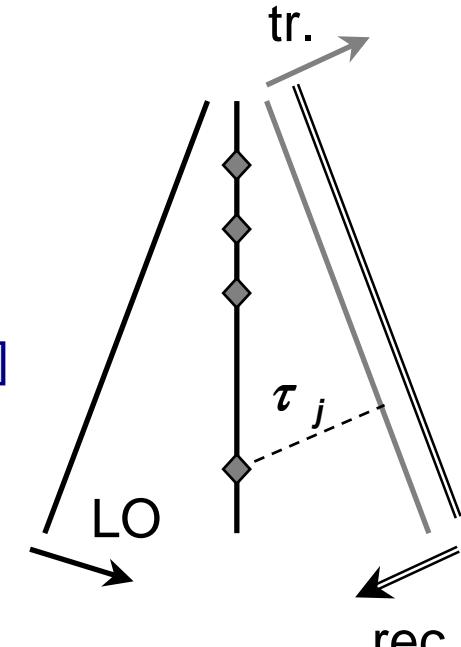
Matched L.O architecture





on channel j :

- transmitted signal : $\cos [2\pi f_{tr} (t - \tau_j)]$
- local oscillator : $\cos [2\pi f_{LO} (t - (\tau_{max} - \tau_j))]$
- received signal : $\cos [2\pi f_r (t - T + \tau_j)]$
- intermediate frequency f_i signal :
 - ▶ $\cos [2\pi (f_{LO} - f_r) t + \underline{2\pi (f_{LO} - f_r) \tau_j} - 2\pi f_{LO} \tau_{max} + 2\pi f_r T]$

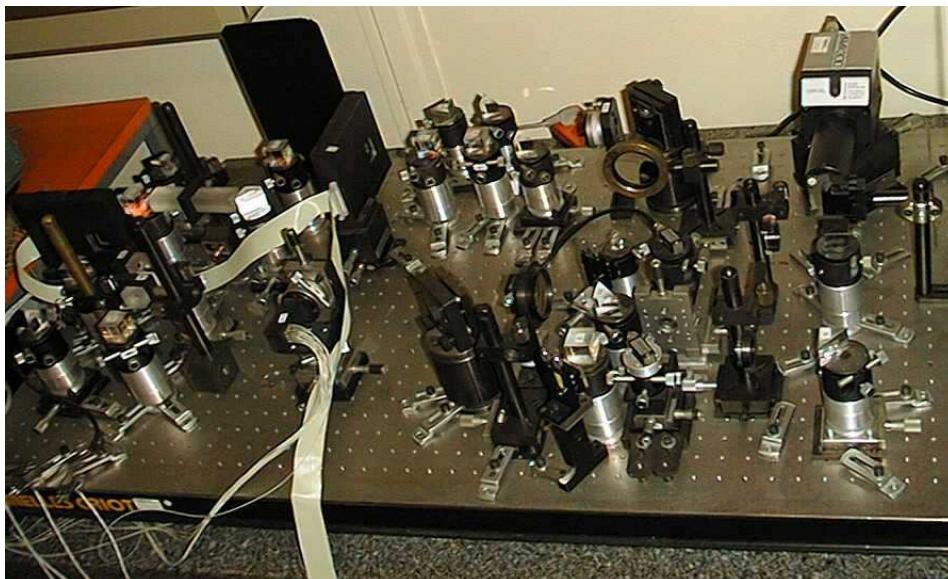


in-phase addition of the signals at f_i when $2\pi (f_{LO} - f_r) \tau_j \ll$ phase quantization

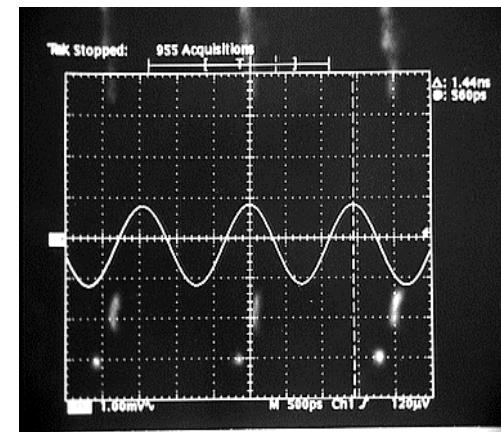


Two channel architecture with:

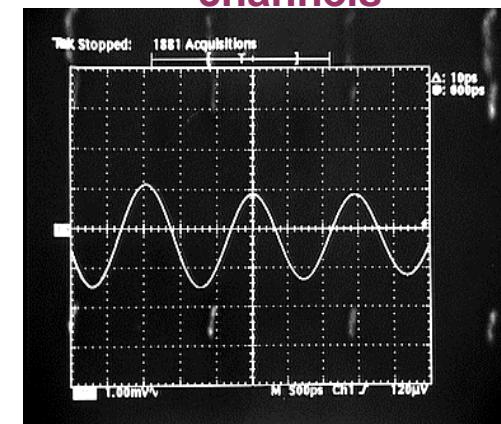
- $f = 2.8 \text{ GHz}$
- I.F= 700 MHz



→in phase I.F signals
→residual errors $\Delta\tau < 10 \text{ ps}$, $\Delta\phi \sim 5^\circ$



I.F signal for no delay
between the
channels



I.F signal for time delay $\tau = 450 \text{ ps}$ between the channels

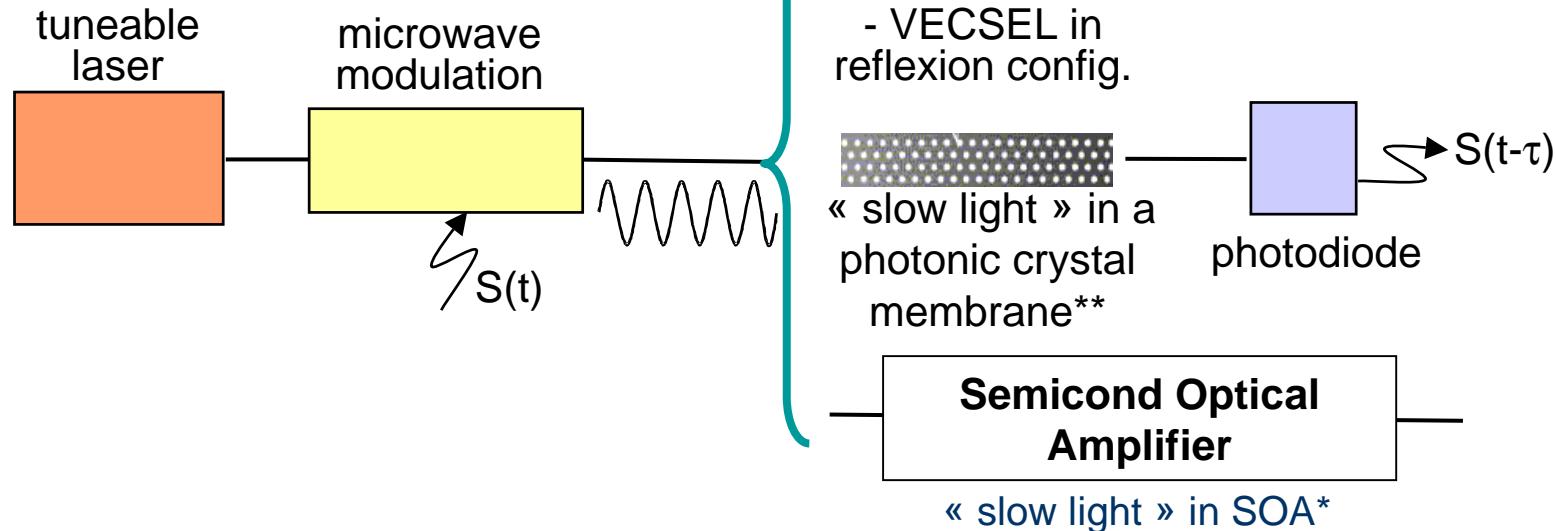
Transmission and optical processing of RF signals



+ Need of variable delay lines for basic processing functions:

- beam control of phased array antennas, programmable filtering
- spectral analysis of large BW microwave signals
- radar phase conjugation

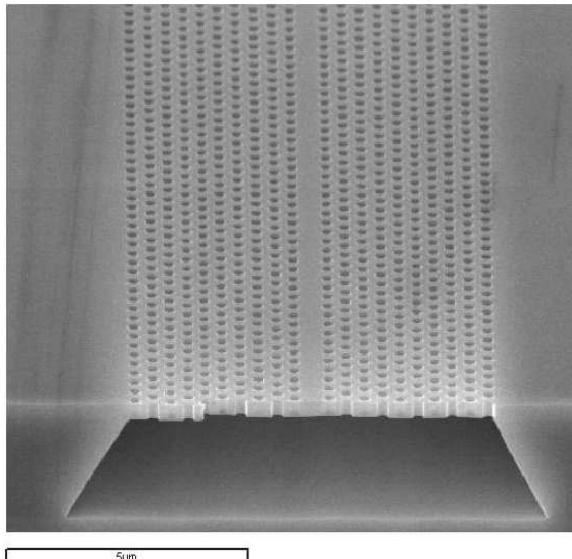
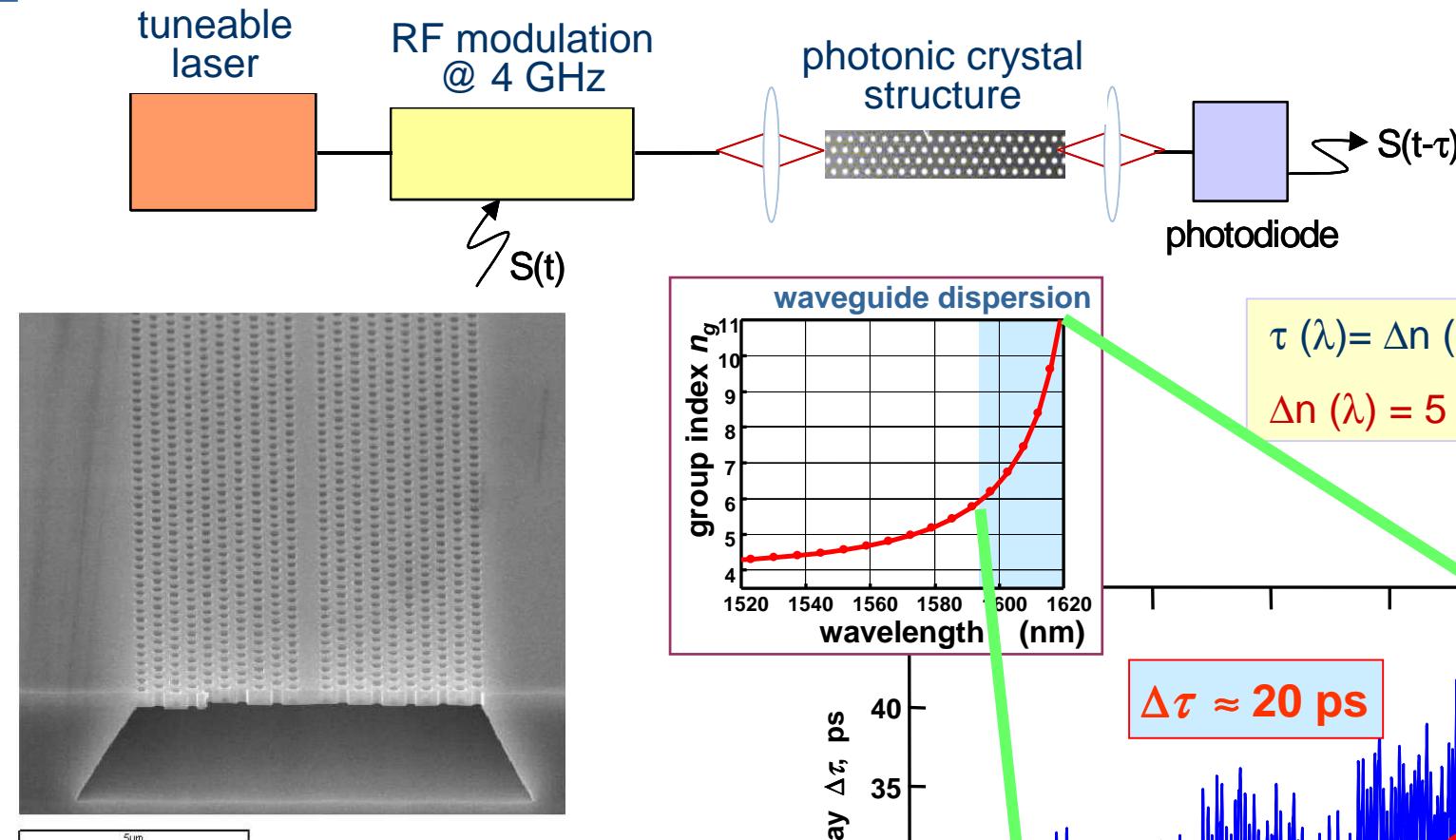
→ **analog optoelectronic link**



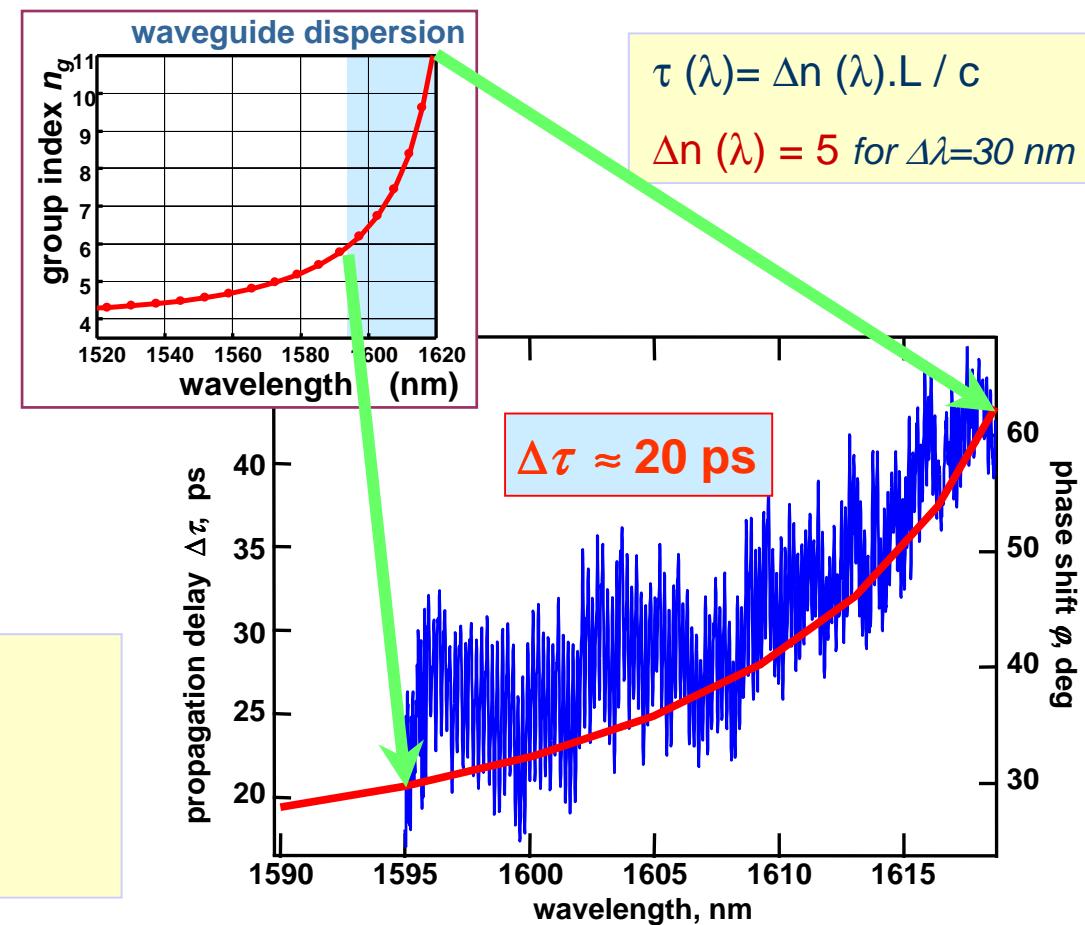
→ *the signal of interest is the microwave modulation itself, not the envelope*

* D. Pastor *et al.*, Electron. Lett., 34, 1684, 1998 ** S. Combrié *et al.*, Electron. Lett., 42, 86, 2006 *** J. Mørk *et al.*, Optics Express., 13, 20, 2005

Photonic Crystal tunable delay preliminary demonstration



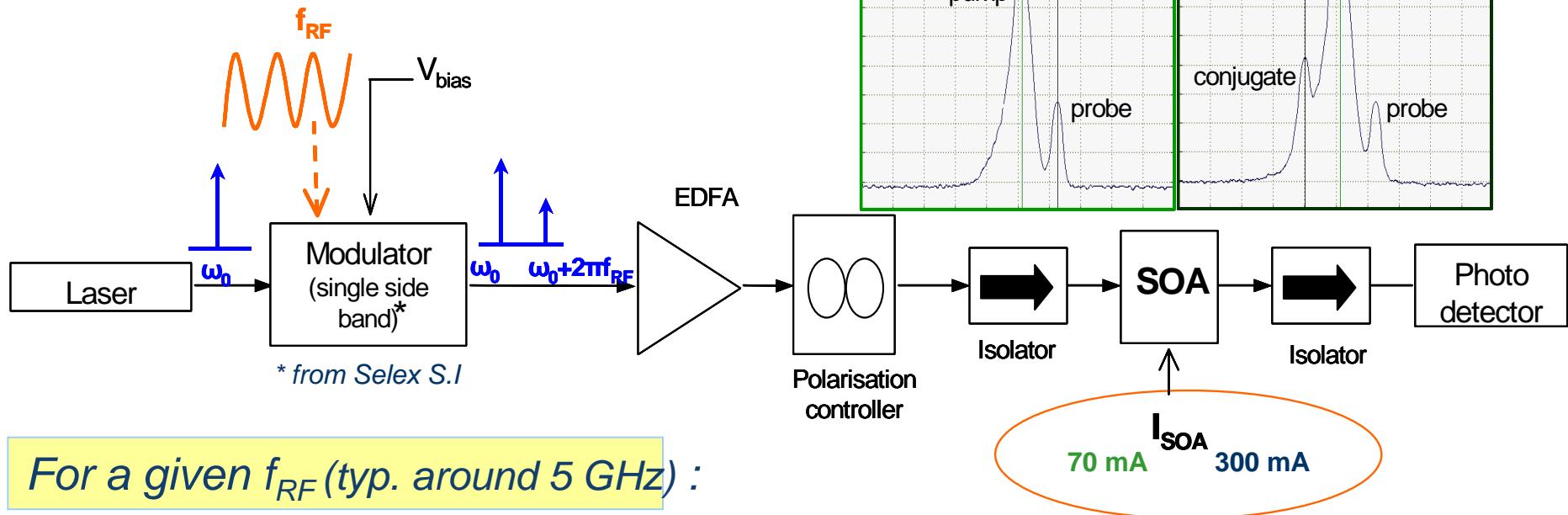
- device length: $L = 1 \text{ mm}$
- measured phase shift: 30 deg @4GHz
- corresp. time delay: $\tau = 21 \text{ ps}$



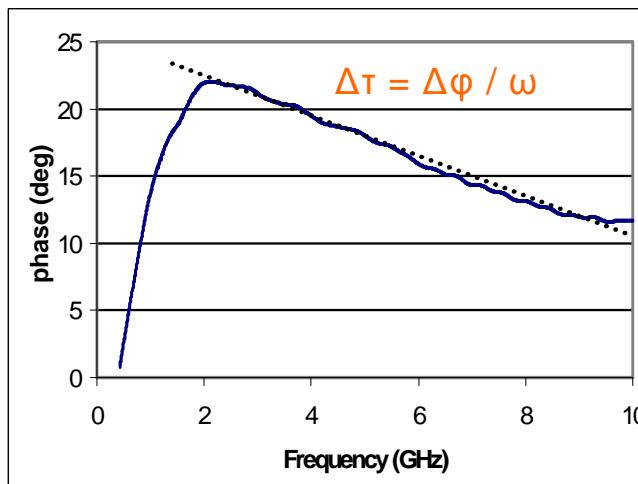
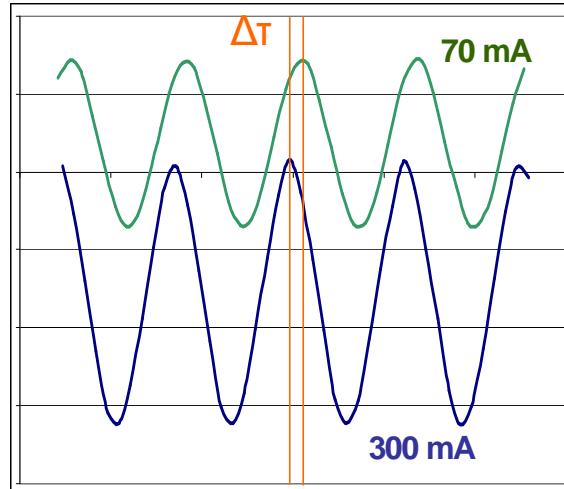
Measurement of the time delays generated with a SOA



Co-propagative set-up



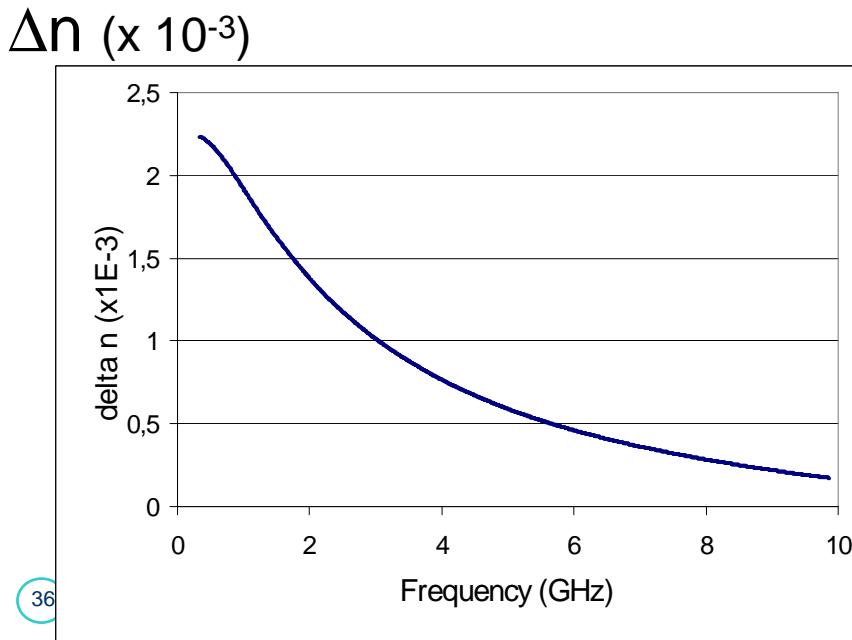
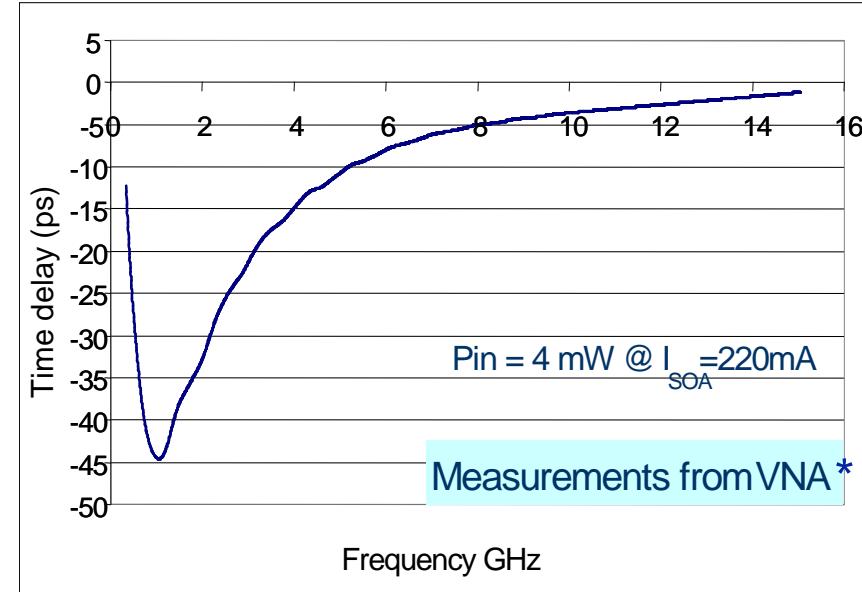
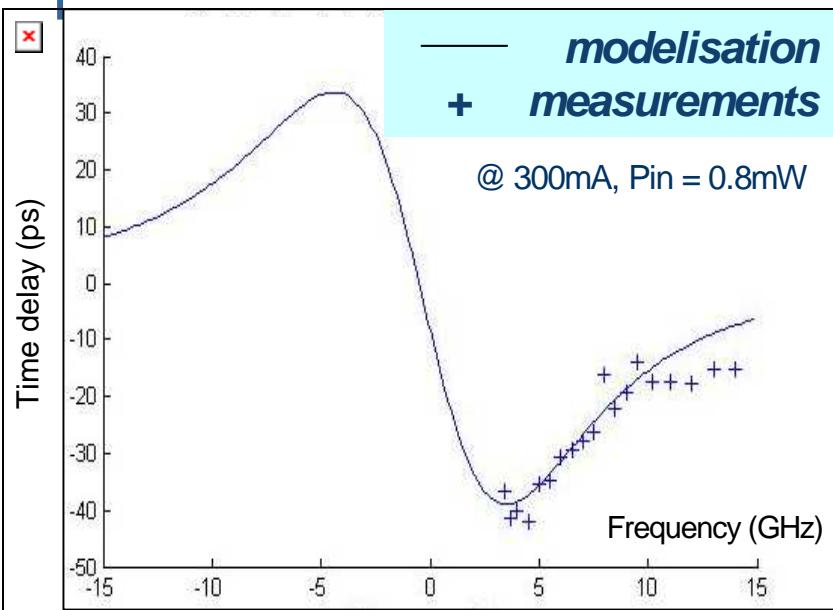
For a given f_{RF} (typ. around 5 GHz) :



$$\Delta n_g = \frac{c}{L} \Delta \tau_g$$

$$\Delta n = \frac{c}{L \omega_{optique}} \int \Delta \tau_g d\Omega$$

Measurement of the time delays generated with a SOA

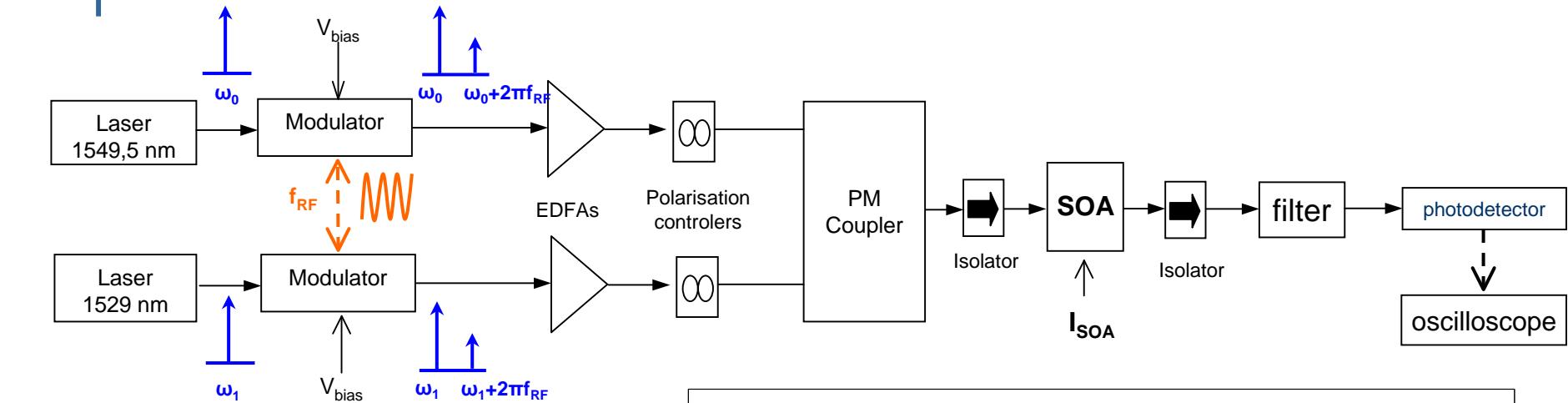


- Time delay from 4 to 45 ps
- Bandwidth : more than 10 GHz
- “Slowlight” with a factor from 2 to 20
- Index variation : few 10^{-3} (through $L=750\text{ }\mu\text{m}$)

*VNA: vectorial network analyzer

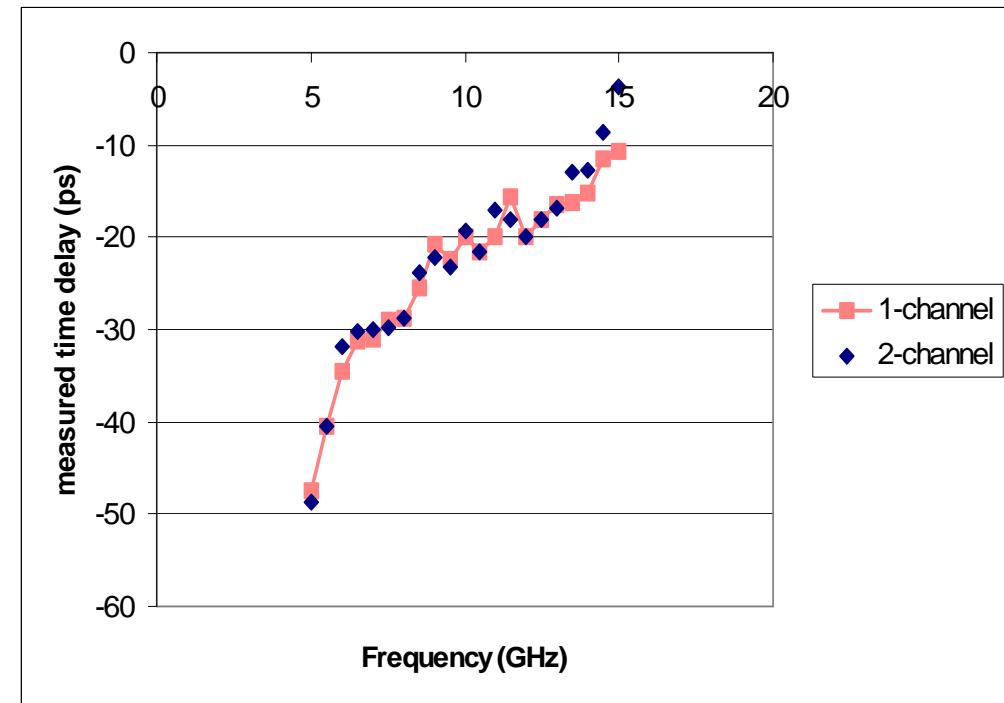
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2-channel experimental demonstration

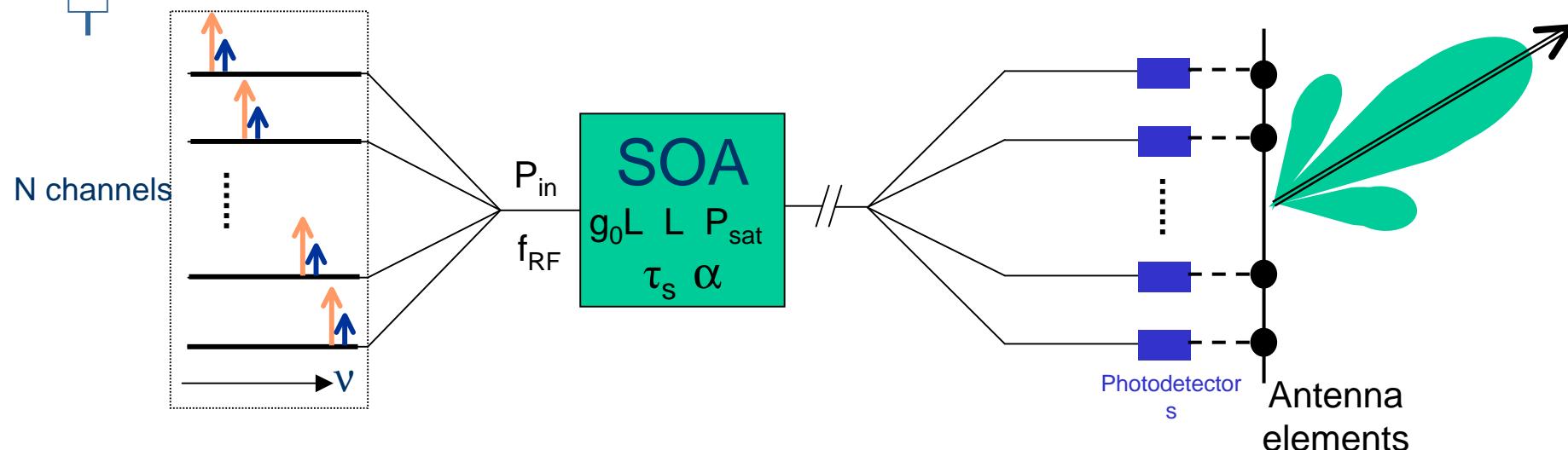


same time delay measured
for the 1-channel
configuration and for the 2-
channel config.*

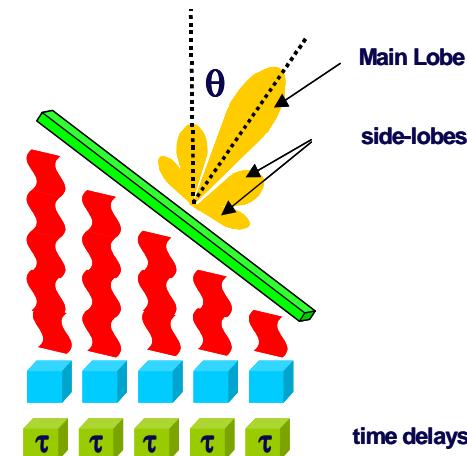
*for one channel present in the opto-link ($P_{in} = 2\text{mW}$) or
for two channels ($P_{in} = 1\text{mW}$ for each)



Applicability to multi-channel phased array antennas

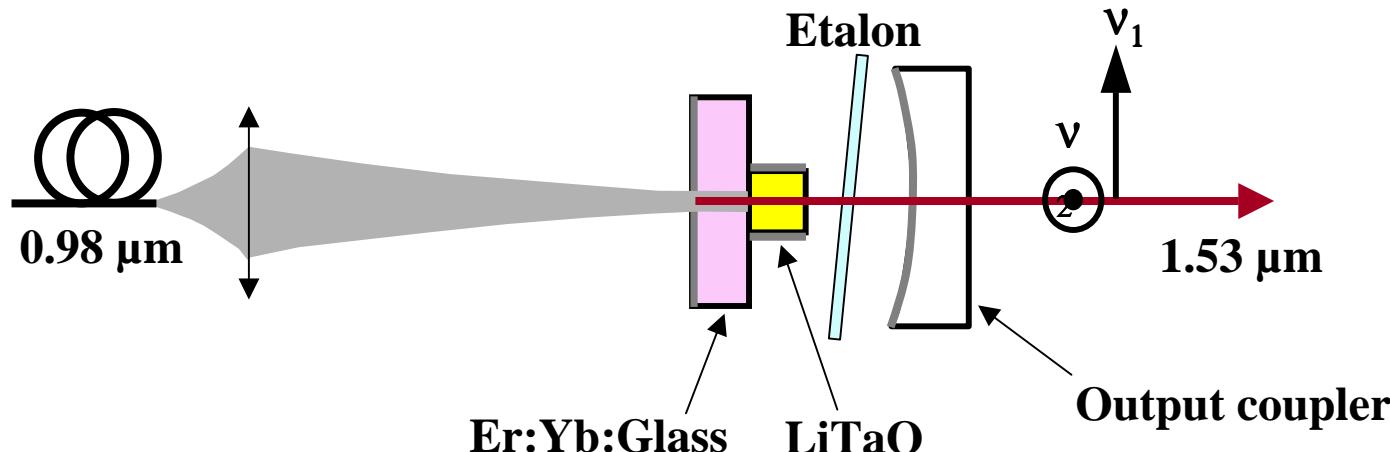


- unchanged SOA characteristics : L τ_s α
- g_0L : homogeneous gain but depending on λ
- $P_{sat} = f(P_{in \text{ total}})$ is unchanged ($P_{sat \text{ 1channel}}/N = P_{sat} / N$)



→ each channel will carry the time delay that it would have carried with the same value of P_{in}/P_{sat}

Er:Yb:Glass dual-frequency laser (DFL)



Active medium : 750 μm long Er:Yb:Glass

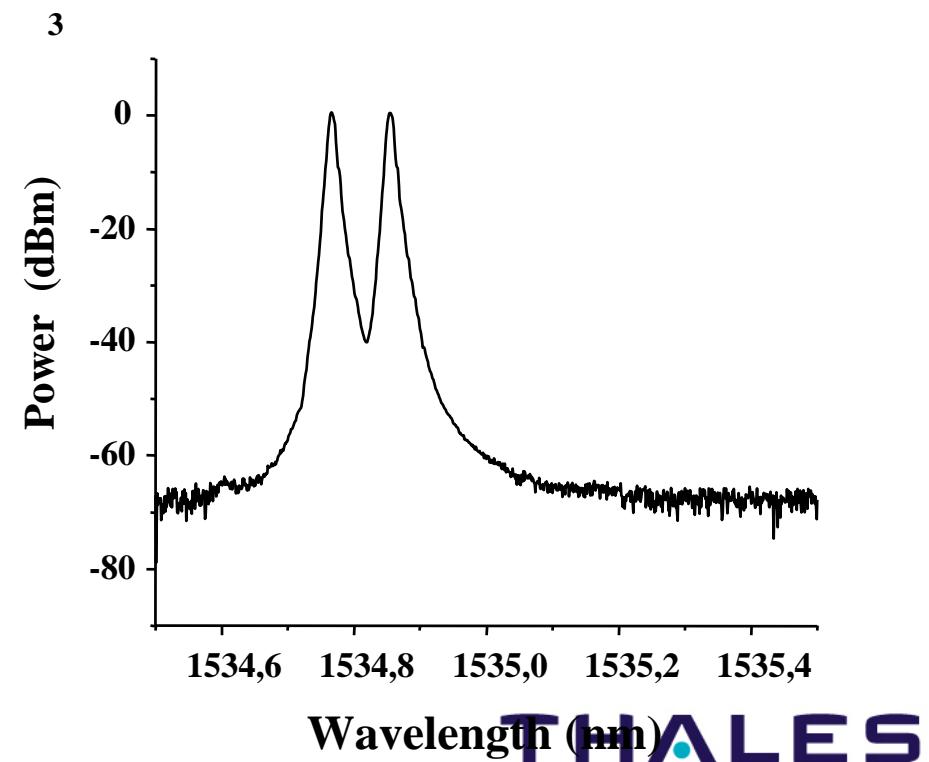
Birefringent element : 1 mm long LiTaO₃ crystal

Pump : single mode diode @ 0.98 μm

Intra-cavity étalon

Frequency difference tunability :

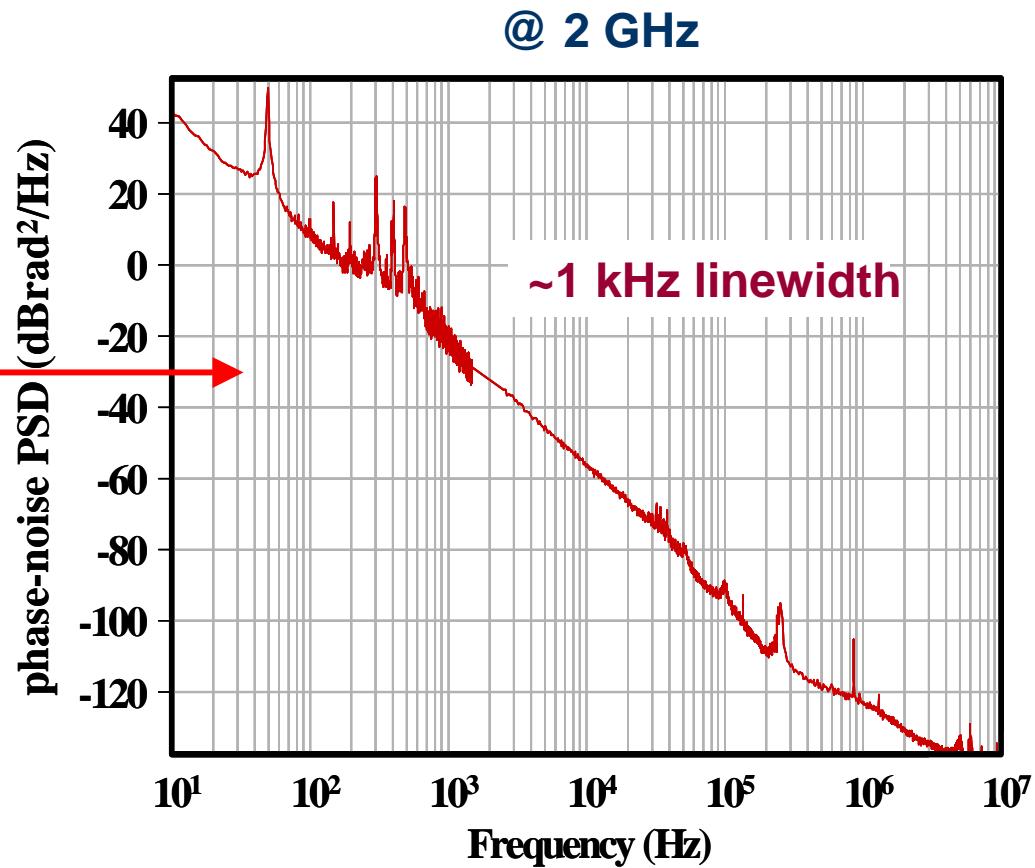
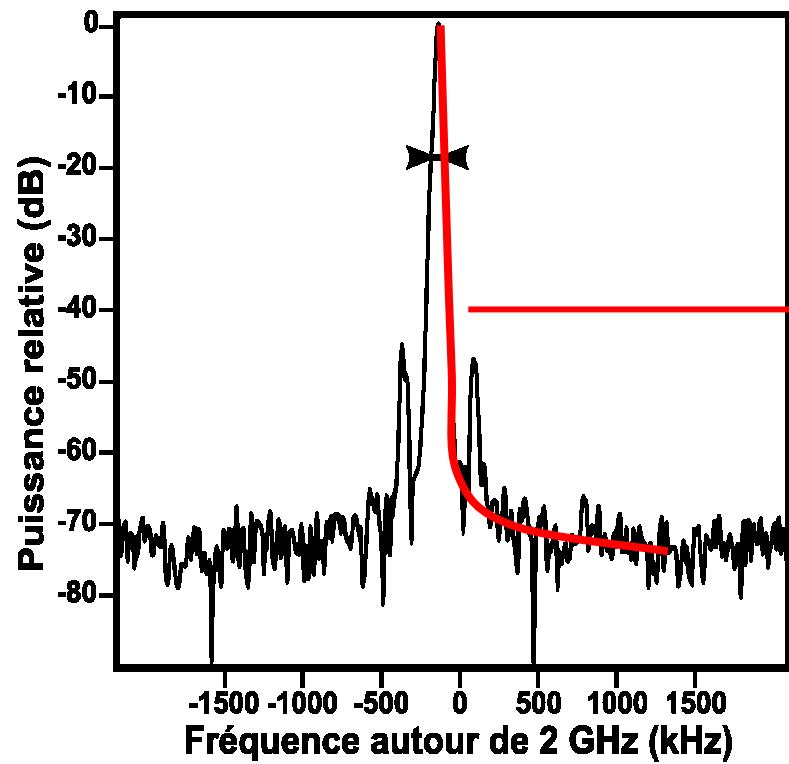
$$\Delta\nu = \Delta\nu_0 + \underbrace{\Gamma V}_{\text{Electro-optic}} + \underbrace{\Lambda(T - T_0)}_{\text{Thermo-optic}}$$



Er:Yb:Glass two-frequency laser (DFL)

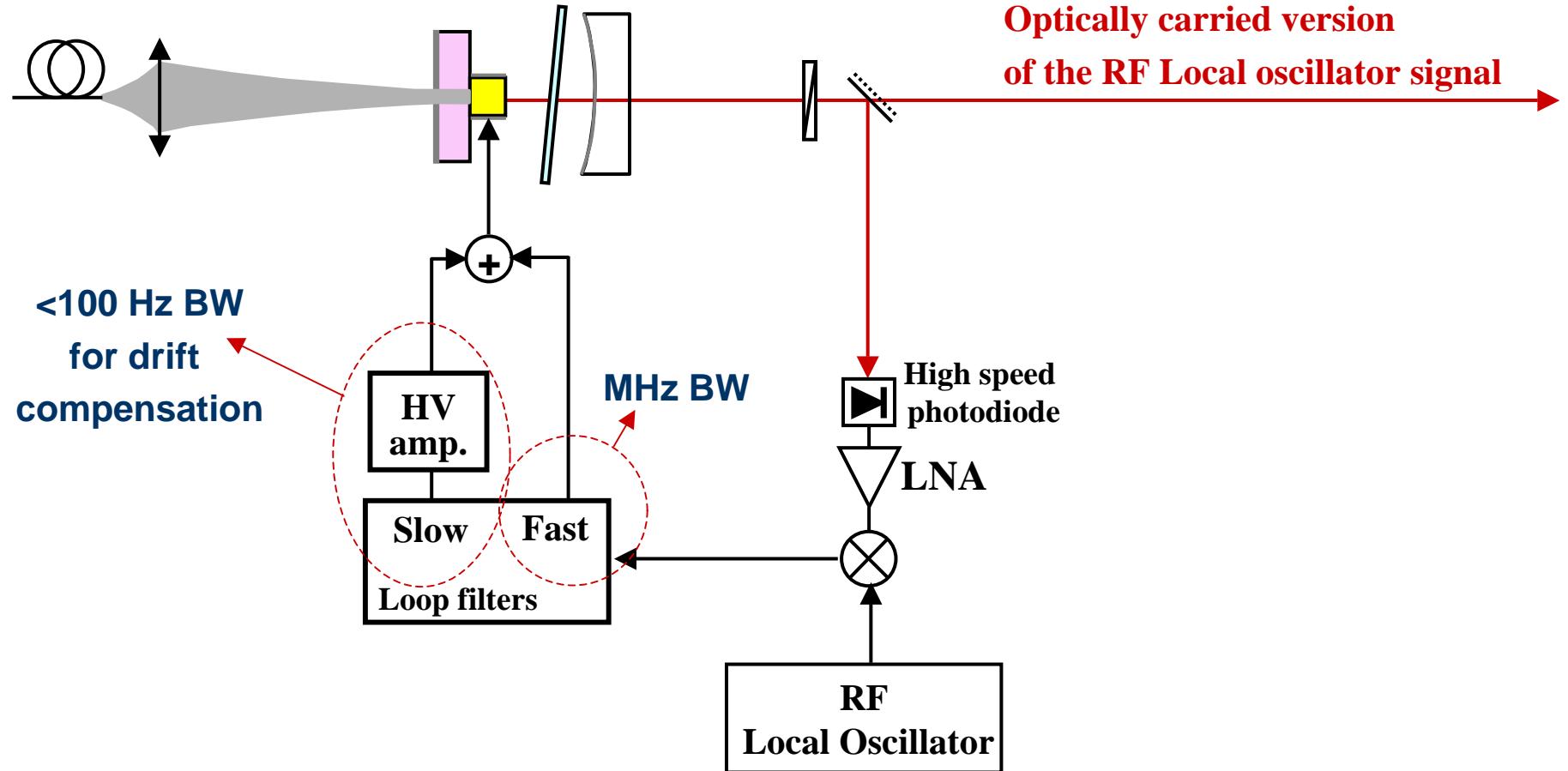


Phase noise of the free-running beatnote



→ Require stabilization for radar applications

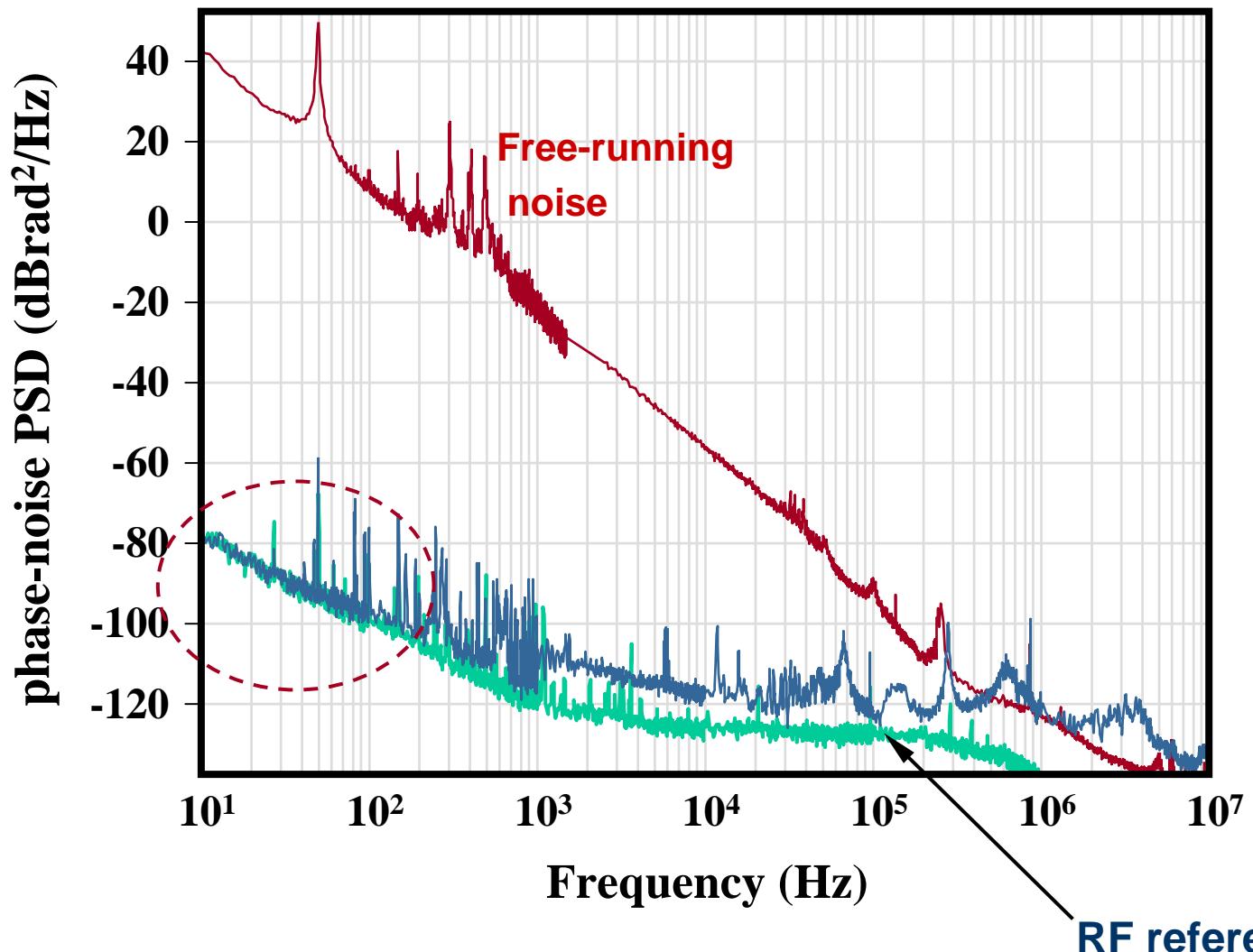
Optical Phase-Locked Loop (OPLL) stabilization



DFL stabilization with an Optical Phase-Locked-Loop (OPLL)



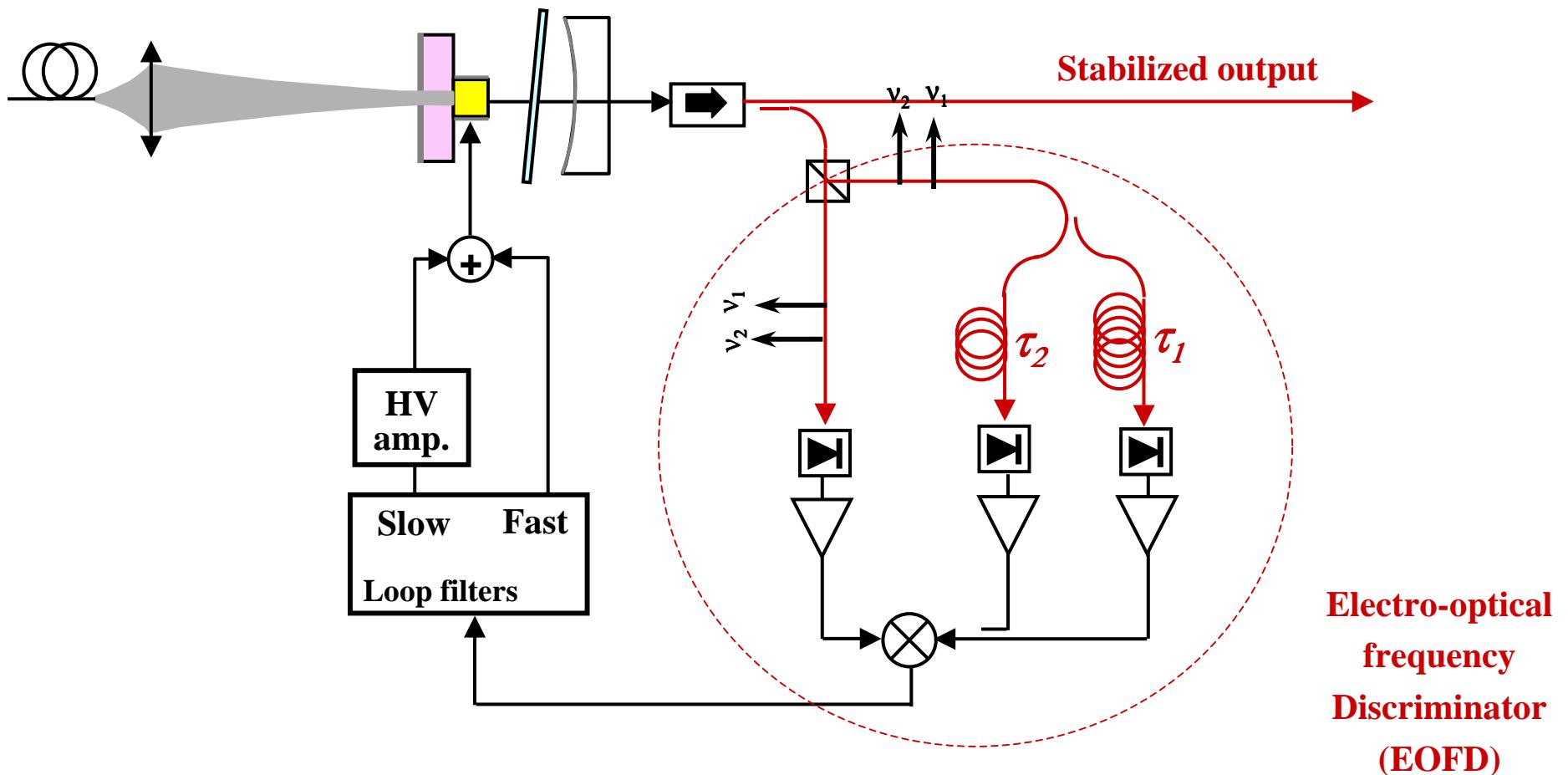
Phase noise of the stabilized beatnote



Fiber delay stabilization



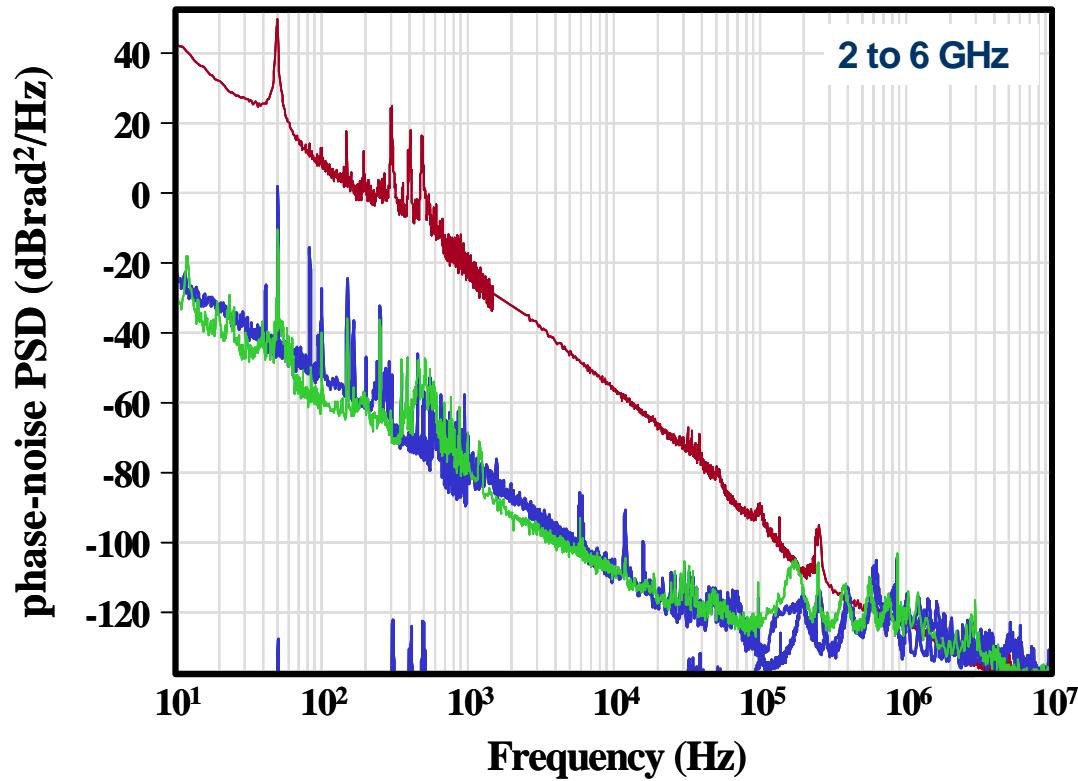
Stabilization combining two delays



Electro-optical
frequency
Discriminator
(EOFD)



Phase noise of the stabilized beatnote : 1000 + 100 meters long fibers



→ both low and high frequency phase noise are reduced, detection + electronic noises are still limiting

- Theoretical limit is fixed by the shot noise level at detection and the amount of delay :
- Low noise, widely tunable, high frequency oscillators for radar applications

$$S_\phi = \frac{2e}{I_{phot} \pi^2 \tau^2 f^2}, \quad f \ll 1/\tau$$

-150 dBBrad^2/Hz @ 10 kHz
(1 km long fiber and 10 mA photocurrent I_{phot})



- **Optical technology offers :**

- ▶ Potential for reducing mass and volume (e.g. the good performance of optical fiber as transmission medium regarding weight, volume and loss can implement compact optically-based TTD beamforming structures).
 - ▶ Antenna remoting
 - ▶ Advance functionality (processing of different types of signals, wideband, multibeam, reconfigurable)
 - ▶ Different approaches to optical beamforming that can be combined to perform the most suitable optical architecture regarding the application targeted
-
- **New approaches for time delay implementation under studies will offer in the future a higher level of integration**